

### MIXED SIGNAL MICROCONTROLLER

### **FEATURES**

- Low Supply Voltage Range: 1.8 V to 3.6 V
- Ultra-Low Power Consumption
  - Active Mode: 250 µA at 1 MHz, 2.2 V
  - Standby Mode: 0.7 μA
  - Off Mode (RAM Retention): 0.1 µA
- Ultra-Fast Wake-Up From Standby Mode in Less Than 1 µs
- 16-Bit RISC Architecture, 62.5-ns Instruction Cycle Time
- Basic Clock Module Configurations
  - Internal Frequencies up to 16 MHz With Four Calibrated Frequencies to ±1%
  - 32-kHz Crystal
  - High-Frequency Crystal up to 16 MHz
  - Resonator
  - External Digital Clock Source
- 16-Bit Timer\_A With Three Capture/Compare Registers
- On-Chip Comparator for Analog Signal Compare Function or Slope Analog-to-Digital (A/D) Conversion
- Brownout Detector

- Serial Onboard Programming, No External Programming Voltage Needed, Programmable Code Protection by Security Fuse
- · Bootstrap Loader
- On Chip Emulation Module
- Family Members:
  - MSP430F2101
    - 1KB + 256B Flash Memory
    - 128B RAM
  - MSP430F2111
    - 2KB + 256B Flash Memory
    - 128B RAM
  - MSP430F2121
    - 4KB + 256B Flash Memory
    - 256B RAM
  - MSP430F2131
    - 8KB + 256B Flash Memory
    - 256B RAM
- Available in a 20-Pin Plastic Small-Outline Wide Body (SOWB) Package, 20-Pin Plastic Small-Outline Thin (TSSOP) Package, 20-Pin TVSOP Package, and 24-Pin QFN Package
- For Complete Module Descriptions, See the MSP430x2xx Family User's Guide (SLAU144)

### **DESCRIPTION**

The Texas Instruments MSP430 family of ultra-low-power microcontrollers consist of several devices featuring different sets of peripherals targeted for various applications. The architecture, combined with five low-power modes, is optimized to achieve extended battery life in portable measurement applications. The device features a powerful 16-bit RISC CPU, 16-bit registers, and constant generators that contribute to maximum code efficiency. The digitally controlled oscillator (DCO) allows wake-up from low-power modes to active mode in less than 1  $\mu$ s.

The MSP430x21x1 series is an ultra-low-power mixed signal microcontroller with a built-in 16-bit timer, versatile analog comparator, and sixteen I/O pins.

Typical applications include sensor systems that capture analog signals, convert themto digital values, and then process the data for display or for transmission to a host system. Stand-alone RF sensor front end is another area of application. The analog comparator provides slope A/D conversion capability.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



### **Table 1. Available Options**

	PACKAGED DEVICES							
T <sub>A</sub>	PLASTIC 20-PIN SOWB (DW)	PLASTIC 20-PIN TSSOP (PW)	PLASTIC 20-PIN TVSOP (DGV)	PLASTIC 24-PIN QFN (RGE)				
	MSP430F2101IDW	MSP430F2101IPW	MSP430F2101IDGV	MSP430F2101IRGE				
1000 / 0500	MSP430F2111IDW	MSP430F2111IPW	MSP430F2111IDGV	MSP430F2111IRGE				
-40°C to 85°C	MSP430F2121IDW	MSP430F2121IPW	MSP430F2121IDGV	MSP430F2121IRGE				
	MSP430F2131IDW	MSP430F2131IPW	MSP430F2131IDGV	MSP430F2131IRGE				
	MSP430F2101TDW	MSP430F2101TPW	MSP430F2101TDGV	MSP430F2101TRGE				
-40°C to 105°C	MSP430F2111TDW	MSP430F2111TPW	MSP430F2111TDGV	MSP430F2111TRGE				
	MSP430F2121TDW	MSP430F2121TPW	MSP430F2121TDGV	MSP430F2121TRGE				
	MSP430F2131TDW	MSP430F2131TPW	MSP430F2131TDGV	MSP430F2131TRGE				

### **Development Tool Support**

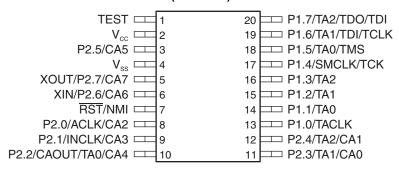
All MSP430 microcontrollers include an Embedded Emulation Module (EEM) that allows advanced debugging and programming through easy-to-use development tools. Recommended hardware options include:

- Debugging and Programming Interface with Target Board
  - MSP-FET430U28 (PW package)
- Debugging and Programming Interface
  - MSP-FET430UIF (USB)
  - MSP-FET430PIF (Parallel Port)
- Target Board
  - MSP-TS430PW28 (PW package)
- Production Programmer
  - MSP-GANG430

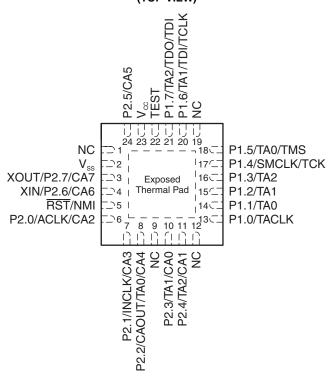


### **Device Pinout**

# DW, PW, or DGV PACKAGE (TOP VIEW)



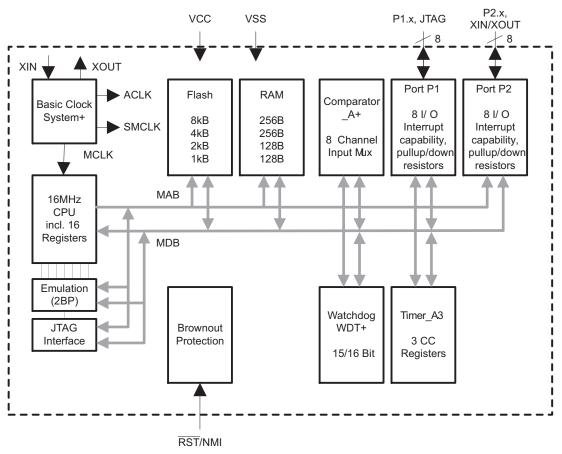
# RGE PACKAGE (TOP VIEW)



- A. NC = Not internally connected
- Exposed thermal pad connection to V<sub>SS</sub> recommended.



### **Functional Block Diagram**



NOTE: See port schematics section for detailed I/O information.



### **Table 2. Terminal Functions**

TERMINAL					
	1	0.			
NAME	DW, PW, or DGV	RGE	I/O	DESCRIPTION	
P1.0/TACLK	13	13	I/O	General-purpose digital I/O pin	
				Timer_A, clock signal TACLK input	
P1.1/TA	14	14	I/O	General-purpose digital I/O pin Timer_A, capture: CCI0A input, compare: Out0 output/BSL transmit	
P1.2/TA1	15	15	I/O	General-purpose digital I/O pin	
				Timer_A, capture: CCI1A input, compare: Out1 output	
P1.3/TA2	16	16	I/O	General-purpose digital I/O pin Timer_A, capture: CCI2A input, compare: Out2 output	
				General-purpose digital I/O pin / SMCLK signal output	
P1.4/SMCLK/TCK	17	17	I/O	Test Clock input for device programming and test	
				General-purpose digital I/O pin / Timer_A, compare: Out0 output	
P1.5/TA/TMS	18	18	I/O	Test Mode Select input for device programming and test	
D				General-purpose digital I/O pin / Timer_A, compare: Out1 output	
P1.6/TA1/TDI/TCLK	19	20	I/O	Test Data Input or Test Clock Input for programming and test	
D4 7/TA 0/TD 0/TD 1(1)	00	0.4	1/0	General-purpose digital I/O pin / Timer_A, compare: Out2 output	
P1.7/TA2/TDO/TDI <sup>(1)</sup>	20	21	I/O	Test Data Output or Test Data Input for programming and test	
P2.0/ACLK/CA2	8	6	I/O	General-purpose digital I/O pin / ACLK output	
FZ.U/AGLIV/GAZ	0	0	1/0	Comparator_A+, CA2 input	
P2.1/INCLK/CA3	9	7	I/O	General-purpose digital I/O pin / Timer_A, clock signal at INCLK	
1 2.1/11(02)(07)(0		'	., 0	Comparator_A+, CA3 input	
				General-purpose digital I/O pin	
P2.2/CAOUT/TA/CA4	10	8	I/O	Timer_A, capture: CCI0B input/BSL receive	
				Comparator_A+, output / CA4 input	
P2.3/CA0/TA1	11	10	I/O	General-purpose digital I/O pin / Timer_A, compare: Out1 output	
				Comparator_A+, CA0 input	
P2.4/CA1/TA2	12	11	I/O	General-purpose digital I/O pin / Timer_A, compare: Out2 output	
				Congrat purpose digital I/O pig	
P2.5/CA5	3	24	I/O	General-purpose digital I/O pin	
				Comparator_A+, CA5 input Input terminal of crystal oscillator	
XIN/P2.6/CA6	6	4	I/O	General-purpose digital I/O pin	
7.11 VI 2.0/0/10		-	.,,	Comparator_A+, CA6 input	
				Output terminal of crystal oscillator	
XOUT/P2.7/CA7 <sup>(2)</sup>	5	3	I/O	·	
				Comparator_A+, CA7 input	
RST/NMI	7	5	I	Reset or nonmaskable interrupt input	
TEST	1	22	I	Selects test mode for JTAG pins on Port1. The device protection fuse is connected to TEST.	
V <sub>CC</sub>	2	23		Supply voltage	
V <sub>SS</sub>	4	2		Ground reference	
QFN Pad	NA	Pad	NA	QFN package thermal pad. Connect to V <sub>SS</sub> .	

<sup>(1)</sup> TDO or TDI is selected via JTAG instruction.

<sup>(2)</sup> If XOUT/P2.7/CA7 is used as an input, excess current will flow until P2SEL.7 is cleared. This is due to the oscillator output driver connection to this pad after reset.



### SHORT-FORM DESCRIPTION

### **CPU**

The MSP430™ CPU has a 16-bit RISC architecture that is highly transparent to the application. All operations, other than program-flow instructions, are performed as register operations in conjunction with seven addressing modes for source operand and four addressing modes for destination operand.

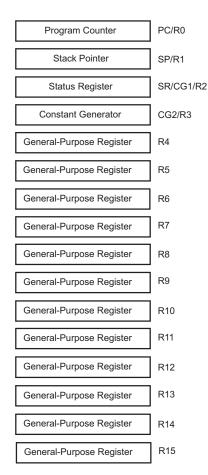
The CPU is integrated with 16 registers that provide reduced instruction execution time. The register-to-register operation execution time is one cycle of the CPU clock.

Four of the registers, R0 to R3, are dedicated as program counter, stack pointer, status register, and constant generator respectively. The remaining registers are general-purpose registers.

Peripherals are connected to the CPU using data, address, and control buses and can be handled with all instructions.

### **Instruction Set**

The instruction set consists of 51 instructions with three formats and seven address modes. Each instruction can operate on word and byte data. Table 3 shows examples of the three types of instruction formats; Table 4 shows the address modes.



**Table 3. Instruction Word Formats** 

INSTRUCTION FORMAT	EXAMPLE	OPERATION
Dual operands, source-destination	ADD R4,R5	R4 + R5 → R5
Single operands, destination only	CALL R8	$PC \rightarrow (TOS), R8 \rightarrow PC$
Relative jump, unconditional/conditional	JNE	Jump-on-equal bit = 0

### **Table 4. Address Mode Descriptions**

ADDRESS MODE	S <sup>(1)</sup>	D <sup>(2)</sup>	SYNTAX	EXAMPLE	OPERATION
Register	✓	✓	MOV Rs,Rd	MOV R10,R11	R10 → R11
Indexed	✓	✓	MOV X(Rn),Y(Rm)	MOV 2(R5),6(R6)	$M(2+R5) \rightarrow M(6+R6)$
Symbolic (PC relative)	✓	✓	MOV EDE,TONI		$M(EDE) \rightarrow M(TONI)$
Absolute	✓	✓	MOV &MEM,&TCDAT		$M(MEM) \rightarrow M(TCDAT)$
Indirect	✓		MOV @Rn,Y(Rm)	MOV @R10,Tab(R6)	$M(R10) \rightarrow M(Tab+R6)$
Indirect autoincrement	✓		MOV @Rn+,Rm	MOV @R10+,R11	M(R10) → R11 R10 + 2 → R10
Immediate	✓		MOV #X,TONI	MOV #45,TONI	#45 → M(TONI)

- (1) S = source
- (2) D = destination



### **Operating Modes**

The MSP430 microcontrollers have one active mode and five software-selectable low-power modes of operation. An interrupt event can wake up the device from any of the five low-power modes, service the request, and restore back to the low-power mode on return from the interrupt program.

The following six operating modes can be configured by software:

- Active mode (AM)
  - All clocks are active.
- Low-power mode 0 (LPM0)
  - CPU is disabled.
  - ACLK and SMCLK remain active. MCLK is disabled.
- Low-power mode 1 (LPM1)
  - CPU is disabled ACLK and SMCLK remain active. MCLK is disabled.
  - DCO dc-generator is disabled if DCO not used in active mode.
- Low-power mode 2 (LPM2)
  - CPU is disabled.
  - MCLK and SMCLK are disabled.
  - DCO dc-generator remains enabled.
  - ACLK remains active.
- Low-power mode 3 (LPM3)
  - CPU is disabled.
  - MCLK and SMCLK are disabled.
  - DCO dc-generator is disabled.
  - ACLK remains active.
- Low-power mode 4 (LPM4)
  - CPU is disabled.
  - ACLK is disabled.
  - MCLK and SMCLK are disabled.
  - DCO dc-generator is disabled.
  - Crystal oscillator is stopped.



### **Interrupt Vector Addresses**

The interrupt vectors and the power-up starting address are located in the address range of 0xFFFF to 0xFFC0. The vector contains the 16-bit address of the appropriate interrupt handler instruction sequence.

If the reset vector (located at address 0xFFFE) contains 0xFFFF (for example, if flash is not programmed), the CPU goes into LPM4 immediately after power up.

**Table 5. Interrupt Vector Addresses** 

INTERRUPT SOURCE	INTERRUPT FLAG	SYSTEM INTERRUPT	WORD ADDRESS	PRIORITY
Power-up	PORIFG			
External reset	RSTIFG			
Watchdog	WDTIFG	Reset	0xFFFE	31, highest
Flash key violation	KEYV <sup>(1)</sup>			
PC out of range <sup>(2)</sup>				
NMI	NMIIFG	(non)-maskable		
Oscillator fault	OFIFG	(non)-maskable	0xFFFC	30
Flash memory access violation	ACCVIFG <sup>(1)(3)</sup>	(non)-maskable		
			0xFFFA	29
			0xFFF8	28
Comparator_A+	CAIFG	maskable	0xFFF6	27
Watchdog Timer+	WDTIFG	maskable	0xFFF4	26
Timer_A3	TACCR0 CCIFG <sup>(4)</sup>	maskable	0xFFF2	25
Timer_A3	TACCR2, TACCR1 CCIFG, TAIFG <sup>(1)(4)</sup>	maskable	0xFFF0	24
			0xFFEE	23
			0xFFEC	22
			0xFFEA	21
			0xFFE8	20
I/O port P2 (eight flags)	P2IFG.0 to P2IFG.7 <sup>(1)(4)</sup>	maskable	0xFFE6	19
I/O port P1 (eight flags)	P1IFG.0 to P1IFG.7 <sup>(1)(4)</sup>	maskable	0xFFE4	18
			0xFFE2	17
			0xFFE0	16
See (5)			0xFFDE	15
See (6)			0xFFDC to 0xFFC0	14 to 0, lowest

<sup>(1)</sup> Multiple source flags

<sup>(2)</sup> A reset is generated if the CPU tries to fetch instructions from within the module register memory address range (0x0000 to 0x01FF) or from within unused address range.

<sup>(3) (</sup>non)-maskable: the individual interrupt-enable bit can disable an interrupt event, but the general interrupt enable cannot.

<sup>4)</sup> Interrupt flags are located in the module.

<sup>(5)</sup> This location is used as bootstrap loader security key (BSLSKEY). A value of 0xAA55 at this location disables the BSL completely.

A value of 0x0 disables the erasure of the flash if an invalid password is supplied.

<sup>(6)</sup> The interrupt vectors at addresses 0xFFDC to 0xFFC0 are not used in this device and can be used for regular program code if necessary.



### **Special Function Registers**

Most interrupt and module enable bits are collected into the lowest address space. Special function register bits not allocated to a functional purpose are not physically present in the device. Simple software access is provided with this arrangement.

### Legend

rw Bit can be read and written.

rw-0, 1 Bit can be read and written. It is Reset or Set by PUC. rw-(0), (1) Bit can be read and written. It is Reset or Set by POR.

SFR bit is not present in device.

### Table 6. Interrupt Enable 1

Address	7	6	5	4	3	2	1	0	
00h			ACCVIE	NMIIE			OFIE	WDTIE	
			rw-0	rw-0			rw-0	rw-0	

WDTIE Watchdog timer interrupt enable. Inactive if watchdog mode is selected. Active if watchdog timer is configured in interval

timer mode.

OFIE Oscillator fault interrupt enable

NMIIE (Non)maskable interrupt enable

ACCVIE Flash access violation interrupt enable

### Table 7. Interrupt Enable 2

Address	7	6	5	4	3	2	1	0
01h								

### Table 8. Interrupt Flag Register 1

Address	7	6	5	4	3	2	1	0
02h				NMIIFG	RSTIFG	PORIFG	OFIFG	WDTIFG
				rw-0	rw-(0)	rw-(1)	rw-1	rw-(0)

WDTIFG Set on watchdog timer overflow (in watchdog mode) or security key violation.

Reset on  $V_{CC}$  power-up or a reset condition at  $\overline{RST}/NMI$  pin in reset mode.

OFIFG Flag set on oscillator fault

RSTIFG External reset interrupt flag. Set on a reset condition at RST/NMI pin in reset mode. Reset on V<sub>CC</sub> power up.

PORIFG Power-on reset interrupt flag. Set on V<sub>CC</sub> power up.

NMIIFG Set via RST/NMI pin

### Table 9. Interrupt Flag Register 2

Address	7	6	5	4	3	2	1	0
03h								



### **Memory Organization**

### **Table 10. Memory Organization**

		MSP430F2101	MSP430F2111	MSP430F2121	MSP430F2131
Memory	Size	1 KB Flash	2 KB Flash	4 KB Flash	8 KB Flash
Main: interrupt vector	Flash	0xFFFF to 0xFFE0	0xFFFF to 0xFFE0	0xFFFF to 0xFFE0	0xFFFF to 0xFFE0
Main: code memory	Flash	0xFFFF to 0xFC00	0xFFFF to 0xF800	0xFFFF to 0xF000	0xFFFF to 0xE000
Information memory	Size	256 Byte	256 Byte	256 Byte	256 Byte
	Flash	0x10FF to 0x1000	0x10FF to 0x1000	0x10FF to 0x1000	0x10FF to 0x1000
Boot memory	Size	1 KB	1 KB	1 KB	1 KB
	ROM	0x0FFF to 0x0C00	0x0FFF to 0x0C00	0x0FFF to 0x0C00	0x0FFF to 0x0C00
RAM	Size	128 B	128 B	256 Byte	256 Byte
		0x027F to 0x0200	0x027F to 0x0200	0x02FF to 0x0200	0x02FF to 0x0200
Peripherals	16-bit	0x01FF to 0x0100	0x01FF to 0x0100	0x01FF to 0x0100	0x01FF to 0x0100
	8-bit	0x0FF to 0x010	0x0FF to 0x010	0x0FF to 0x010	0x0FF to 0x010
	8-bit SFR	0x0F to 0x00	0x0F to 0x00	0x0F to 0x00	0x0F to 0x00

### **Bootstrap Loader (BSL)**

The MSP430 bootstrap loader (BSL) enables users to program the flash memory or RAM using a UART serial interface. Access to the MSP430 memory via the BSL is protected by user-defined password. Abootstrap loader security key is provided at address 0FFDEh to disable the BSL completely or to disable the erasure of the flash if an invalid password is supplied. For complete description of the features of the BSL and its implementation, see the MSP430 Programming Via the Bootstrap Loader User's Guide, literature number SLAU319.

Table 11. BSL Keys

BSLKEY	DESCRIPTION
00000h	Erasure of flash disabled if an invalid password is supplied
0AA55h	BSL disabled
any other value	BSL enabled

**Table 12. BSL Function Pins** 

BSL FUNCTION	DW, PW, DGV PACKAGE PINS	RGE PACKAGE PINS
Data transmit	14 - P1.1	14 - P1.1
Data receive	10 - P2.2	8 - P2.2

### **Flash Memory**

The flash memory can be programmed via the JTAG port, the bootstrap loader, or in-system by the CPU. The CPU can perform single-byte and single-word writes to the flash memory. Features of the flash memory include:

- Flash memory has n segments of main memory and four segments of information memory (A to D) of 64 bytes each. Each segment in main memory is 512 bytes in size.
- Segments 0 to n may be erased in one step, or each segment may be individually erased.
- Segments A to D can be erased individually, or as a group with segments 0 to n.
   Segments A to D are also called information memory.
- Segment A contains calibration data. After reset, segment A is protected against programming and erasing. It
  can be unlocked, but care should be taken not to erase this segment if the device-specific calibration data is
  required.



### **Peripherals**

Peripherals are connected to the CPU through data, address, and control buses and can be handled using all instructions. For complete module descriptions, see the MSP430x2xx Family User's Guide (SLAU144).

### **Oscillator and System Clock**

The clock system is supported by the basic clock module that includes support for a 32768-Hz watch crystal oscillator, an internal digitally-controlled oscillator (DCO), and a high-frequency crystal oscillator. The basic clock module is designed to meet the requirements of both low system cost and low power consumption. The internal DCO provides a fast turn-on clock source and stabilizes in less than 1 µs. The basic clock module provides the following clock signals:

- Auxiliary clock (ACLK), sourced from a 32768-Hz watch crystal or a high-frequency crystal
- Main clock (MCLK), the system clock used by the CPU
- Sub-Main clock (SMCLK), the sub-system clock used by the peripheral modules

Table 13. DCO Calibration Data, Provided From Factory In Flash Info Memory Segment A

DCO FREQUENCY	CALIBRATION REGISTER	SIZE	ADDRESS
4 1411-	CALBC1_1MHZ	byte	0x010FF
1 MHz	CALBC0_1MHZ	byte	0x010FE
8 MHz	CALBC1_8MHZ	byte	0x010FD
O IVITZ	CALBC0_8MHZ	byte	0x010FC
12 MHz	CALBC1_12MHZ	byte	0x010FB
IZ IVITZ	CALBC0_12MHZ	byte	0x010FA
16 MHz	CALBC1_16MHZ	byte	0x010F9
I O IVITZ	CALBC0_16MHZ	byte	0x010F8

### **Brownout**

The brownout circuit is implemented to provide the proper internal reset signal to the device during power on and power off.

### Digital I/O

There are two 8-bit I/O ports implemented—ports P1 and P2.

- All individual I/O bits are independently programmable.
- · Any combination of input, output, and interrupt condition is possible.
- Edge-selectable interrupt input capability for all eight bits of port P1 and P2.
- Read/write access to port-control registers is supported by all instructions.
- Each I/O has an individually programmable pullup/pulldown resistor.

### Watchdog Timer (WDT+)

The primary function of the WDT+ module is to perform a controlled system restart after a software problem occurs. If the selected time interval expires, a system reset is generated. If the watchdog function is not needed in an application, the module can be disabled or configured as an interval timer and can generate interrupts at selected time intervals.

### Comparator\_A+

The primary function of the comparator\_A+ module is to support precision slope analog-to-digital conversions, battery-voltage supervision, and monitoring of external analog signals.



### Timer\_A3

Timer\_A3 is a 16-bit timer/counter with three capture/compare registers. Timer\_A3 can support multiple capture/compares, PWM outputs, and interval timing. Timer\_A3 also has extensive interrupt capabilities. Interrupts may be generated from the counter on overflow conditions and from each of the capture/compare registers.

Table 14. Timer\_A3 Signal Connections

INPUT PIN	NUMBER	DEVICE INPUT	MODULE	MODULE	MODULE	OUTPUT PIN	NUMBER
DW, PW, DGV	RGE	SIGNAL	INPUT NAME	BLOCK	OUTPUT SIGNAL	DW, PW, DGV	RGE
13 - P1.0	13 - P1.0	TACLK	TACLK				
		ACLK	ACLK	<del>-</del>	NA		
		SMCLK	SMCLK	Timer			
9 - P2.1	7 - P2.1	INCLK	INCLK				
14 - P1.1	14 - P1.1	TA	CCI0A			14 - P1.1	14 - P1.1
10 - P2.2	8 - P2.2	TA	CCI0B	0000	Τ.	18 - P1.5	18 - P1.5
		VSS	GND	CCR0	TA		
		VCC	VCC				
15 - P1.2	15 - P1.2	TA1	CCI1A			11 - P2.3	10 - P2.3
		CAOUT (internal)	CCI1B	CCR1	TA1	15 - P1.2	15 - P1.2
		VSS	GND			19 - P1.6	20 - P1.6
		VCC	VCC				
16 - P1.3	16 - P1.3	TA2	CCI2A			12 - P2.4	11 - P2.4
		ACLK (internal)	CCI2B	CCDO	TA0	16 - P1.3	16 - P1.3
		VSS	GND	CCR2	TA2	20 - P1.7	21 - P1.7
		VCC	VCC				



## **Peripheral File Map**

### **Table 15. Peripherals With Word Access**

MODULE	REGISTER NAME	SHORT NAME	ADDRESS OFFSET
Timer_A	Capture/compare register	TACCR2	0x0176
	Capture/compare register	TACCR1	0x0174
	Capture/compare register	TACCR0	0x0172
	Timer_A3 register	TAR	0x0170
	Capture/compare control	TACCTL2	0x0166
	Capture/compare control	TACCTL1	0x0164
	Capture/compare control	TACCTL0	0x0162
	Timer_A3 control	TACTL	0x0160
	Timer_A3 interrupt vector	TAIV	0x012E
Flash Memory	Flash control 3	FCTL3	0x012C
	Flash control 2	FCTL2	0x012A
	Flash control 1	FCTL1	0x0128
Watchdog Timer+	Watchdog/timer control	WDTCTL	0x0120

### **Table 16. Peripherals With Byte Access**

MODULE	REGISTER NAME	SHORT NAME	ADDRESS OFFSET
Comparator_A+	Comparator_A port disable	CAPD	0x005B
	Comparator_A control 2	CACTL2	0x005A
	Comparator_A control 1	CACTL1	0x0059
Basic Clock	Basic clock system control 3	BCSCTL3	0x0053
	Basic clock system control 2	BCSCTL2	0x0058
	Basic clock system control 1	BCSCTL1	0x0057
	DCO clock frequency control	DCOCTL	0x0056
Port P2	Port P2 resistor enable	P2REN	0x002F
	Port P2 selection	P2SEL	0x002E
	Port P2 interrupt enable	P2IE	0x002D
	Port P2 interrupt edge select	P2IES	0x002C
	Port P2 interrupt flag	P2IFG	0x002B
	Port P2 direction	P2DIR	0x002A
	Port P2 output	P2OUT	0x0029
	Port P2 input	P2IN	0x0028
Port P1	Port P1 resistor enable	P1REN	0x0027
	Port P1 selection	P1SEL	0x0026
	Port P1 interrupt enable	P1IE	0x0025
	Port P1 interrupt edge select	P1IES	0x0024
	Port P1 interrupt flag	P1IFG	0x0023
	Port P1 direction	P1DIR	0x0022
	Port P1 output	P1OUT	0x0021
	Port P1 input	P1IN	0x0020
Special Function	SFR interrupt flag 2	IFG2	0x0003
	SFR interrupt flag 1	IFG1	0x0002
	SFR interrupt enable 2	IE2	0x0001
	SFR interrupt enable 1	IE1	0x0000



### Absolute Maximum Ratings(1)

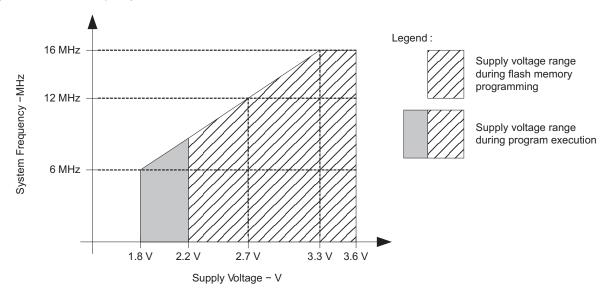
Voltage applied at V <sub>CC</sub> to V <sub>SS</sub>		-0.3 V to 4.1 V
Voltage applied to any pin (2)		-0.3 V to (V <sub>CC</sub> + 0.3 V)
Diode current at any device terminal		±2 mA
Ctorono torrono T (3)	Unprogrammed device	-55°C to 150°C
Storage temperature, T <sub>stg</sub> <sup>(3)</sup>	Programmed device	-55°C to 150°C

- (1) Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltages referenced to V<sub>SS</sub>. The JTAG fuse-blow voltage, V<sub>FB</sub>, is allowed to exceed the absolute maximum rating. The voltage is applied to the TEST pin when blowing the JTAG fuse.
- (3) Higher temperature may be applied during board soldering process according to the current JEDEC J-STD-020 specification with peak reflow temperatures not higher than classified on the device label on the shipping boxes or reels.

### Recommended Operating Conditions<sup>(1)</sup>

			MIN	NOM	MAX	UNIT
V	Supply voltage, AV <sub>CC</sub> = DV <sub>CC</sub> = V <sub>CC</sub> During program execution  During flash memory programming	During program execution	1.8		3.6	V
V <sub>CC</sub>		During flash memory programming	2.2		3.6	V
$V_{SS}$	Supply voltage, $AV_{SS} = DV_{SS} = V_{SS}$			0		V
_	Operating free-air temperature	I version	-40		85	°C
T <sub>A</sub>		T version	-40		105	C
	Processor frequency (maximum MCLK	$V_{CC} = 1.8 \text{ V}$ , Duty cycle = 50% ±10%	0		6	
f <sub>SYSTEM</sub>	Processor frequency (maximum MCLK frequency) (2)(1)	$V_{CC} = 2.7 \text{ V}$ , Duty cycle = 50% ±10%	0		12	MHz
	(see Figure 1)	$V_{CC} \ge 3.3 \text{ V}$ , Duty cycle = 50% ±10%	0		16	

- (1) Modules might have a different maximum input clock specification. See the specification of the respective module in this data sheet.
- (2) The MSP430 CPU is clocked directly with MCLK. Both the high and low phase of MCLK must not exceed the pulse width of the specified maximum frequency.



NOTE: Minimum processor frequency is defined by system clock. Flash program or erase operations require a minimum V<sub>CC</sub> of 2.2 V.

Figure 1. Operating Area



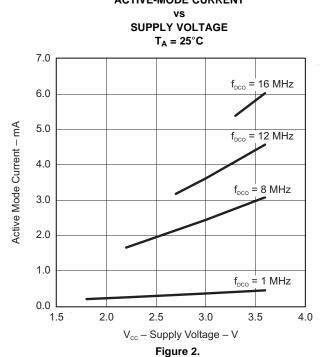
### Active Mode Supply Current (into DV<sub>cc</sub> + AV<sub>cc</sub> ) Excluding External Current

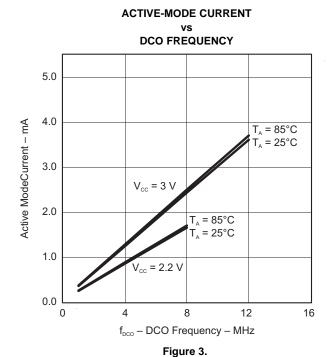
over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)(1)(2)

F	PARAMETER	TEST CONDITIONS	$T_A$	V <sub>cc</sub>	MIN TYP	MAX	UNIT	
I <sub>AM,1MHz</sub>	Active mode (AM) current (1 MHz)	$\begin{split} f_{DCO} &= f_{MCLK} = f_{SMCLK} = 1 \text{ MHz}, \\ f_{ACLK} &= 32768 \text{ Hz}, \\ Program executes in flash, \\ BCSCTL1 &= CALBC1\_1MHZ, \\ DCOCTL &= CALDCO\_1MHZ, \\ CPUOFF &= 0, SCG0 = 0, SCG1 = 0, \\ OSCOFF &= 0 \end{split}$		2.2 V 3 V	250 350	300 410	μΑ	
		$f_{DCO} = f_{MCLK} = f_{SMCLK} = 1 \text{ MHz},$		2.2 V	200			
I <sub>AM,1MHz</sub>	Active mode (AM) current (1 MHz)	f <sub>ACLK</sub> = 32768 Hz, Program executes in RAM, BCSCTL1 = CALBC1_1MHZ, DCOCTL = CALDCO_1MHZ, CPUOFF = 0, SCG0 = 0, SCG1 = 0, OSCOFF = 0		3 V	300		μΑ	
		$f_{MCLK} = f_{SMCLK} = f_{ACLK} = 32768 \text{ Hz} / 8$	-40°C to 85°C	221/	2.2 V	2	5	
		= 4096 Hz, f <sub>DCO</sub> = 0 Hz,	105°C	Z.Z V		6		
lana arar-	Active mode (AM)	Program executes in flash,	-40°C to 85°C		3	9	μA	
IAM,4kHz	current (4 kHz)	SELMx = 11, SELS = 1, DIVMx = DIVSx = DIVAx = 11, CPUOFF = 0, SCG0 = 1, SCG1 = 0, OSCOFF = 0	105°C	3 V		9	μπ	
		$f_{MCLK} = f_{SMCLK} = f_{DCO(0, 0)} \approx 100 \text{ kHz},$		2.2 V	60	85		
I <sub>AM,100kHz</sub>	Active mode (AM) current (100 kHz)	f <sub>ACLK</sub> = 0 Hz, Program executes in flash, RSELx = 0, DCOx = 0, CPUOFF = 0, SCG0 = 0, SCG1 = 0, OSCOFF = 1		3 V	72	95	μΑ	

- All inputs are tied to 0 V or  $V_{CC}$ . Outputs do not source or sink any current. The currents are characterized with a Micro Crystal CC4V-T1A SMD crystal with a load capacitance of 9 pF. The internal and external load capacitance is chosen to closely match the required 9 pF.

### Typical Characteristics - Active-Mode Supply Current (Into V<sub>CC</sub>) **ACTIVE-MODE CURRENT**







# Low-Power-Mode Supply Currents (Into $V_{\text{CC}}$ ) Excluding External Current $^{(1)(2)}$

P	ARAMETER	TEST CONDITIONS	T <sub>A</sub>	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
I <sub>LPM0,1MHz</sub>	Low-power mode 0 (LPM0) current <sup>(3)</sup>	$\begin{split} &f_{MCLK} = 0 \text{ MHz}, \\ &f_{SMCLK} = f_{DCO} = 1 \text{ MHz}, \\ &f_{ACLK} = 32768 \text{ Hz}, \\ &BCSCTL1 = CALBC1\_1MHZ, \\ &DCOCTL = CALDCO\_1MHZ, \\ &CPUOFF = 1, SCG0 = 0, SCG1 = 0, \\ &OSCOFF = 0 \end{split}$		2.2 V 3 V		65 85	100	μА
		f <sub>MCLK</sub> = 0 MHz,		2.2 V		37	48	
I <sub>LPM0,100kHz</sub>	Low-power mode 0 (LPM0) current <sup>(3)</sup>	$\begin{split} &f_{SMCLK} = f_{DCO}(0,0) \approx 100\text{ kHz},\\ &f_{ACLK} = 0\text{ Hz},\\ &RSELx = 0,DCOx = 0,\\ &CPUOFF = 1,SCG0 = 0,SCG1 = 0,\\ &OSCOFF = 1 \end{split}$		3 V		41	52	μΑ
		$f_{MCLK} = f_{SMCLK} = 0 \text{ MHz},$	-40°C to 85°C	221		22	22 29	
	Low-power mode 2	f <sub>DCO</sub> = 1 MHz, f <sub>ACLK</sub> = 32768 Hz, BCSCTL1 = CALBC1_1MHZ,	105°C	2.2 V			31	
I <sub>LPM2</sub>	(LPM2) current <sup>(4)</sup>	DCOCTL = CALDCO_1MHZ, CPUOFF = 1, SCG0 = 0, SCG1 = 1, OSCOFF = 0	-40°C to 85°C			25	32	μА
			105°C	3 V			34	
			-40°C			0.7	1.2	
			25°C	2.2.7		0.7	1	
		6 6 0 MHz	85°C	2.2 V		1.6	2.3	
	Low-power mode 3	$f_{DCO} = f_{MCLK} = f_{SMCLK} = 0 \text{ MHz},$ $f_{ACLK} = 32768 \text{ Hz},$	105°C			3	6	
I <sub>LPM3,LFXT1</sub>	(LPM3) current <sup>(4)</sup>	CPUOFF = 1, $SCG0 = 1$ , $SCG1 = 1$ ,	-40°C			0.9	1.2	μA
		OSCOFF = 0	25°C	0.1/		0.9	1.2	
			85°C	3 V		1.6	2.8	
			105°C			3	7	
		f - f - f - O ML -	-40°C			0.1	0.5	
	Low-power mode 4	$f_{DCO} = f_{MCLK} = f_{SMCLK} = 0 \text{ MHz},$ $f_{ACLK} = 0 \text{ Hz},$	25°C	2 2 1/2 1/		0.1	0.5	
I <sub>LPM4</sub>	(LPM4) current <sup>(5)</sup>	CPUOFF = 1, SCG0 = 1, SCG1 = 1,	85°C	2.2 V/3 V		8.0	1.9	μA
		OSCOFF = 1	105°C	[		2	4	

 <sup>(1)</sup> All inputs are tied to 0 V or V<sub>CC</sub>. Outputs do not source or sink any current.
 (2) The currents are characterized with a Micro Crystal CC4V-T1A SMD crystal with a load capacitance of 9 pF. The internal and external load capacitance is chosen to closely match the required 9 pF.
 (3) Current for brownout and WDT clocked by SMCLK included.

Current for brownout and WDT clocked by ACLK included.

Current for brownout included.



### Schmitt-Trigger Inputs (Ports P1, P2)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
				0.45 V <sub>CC</sub>		0.75 V <sub>CC</sub>	
$V_{IT+}$	Positive-going input threshold voltage		2.2 V	1		1.65	V
			3 V	1.35		2.25	
	Negative-going input threshold voltage			0.25 V <sub>CC</sub>		0.55 V <sub>CC</sub>	
$V_{IT-}$			2.2 V	0.55		1.20	V
			3 V	0.75		1.65	
\/	Input valtage bysteresis ()/		2.2 V	0.2		1	V
$V_{hys}$	Input voltage hysteresis (V <sub>IT+</sub> - V <sub>IT-</sub> )		3 V	0.3		1	V
R <sub>Pull</sub>	Pullup/pulldown resistor	For pullup: $V_{IN} = V_{SS}$ , For pulldown: $V_{IN} = V_{CC}$		20	35	50	kΩ
Cı	Input capacitance	$V_{IN} = V_{SS}$ or $V_{CC}$			5		pF

### Inputs (Ports P1, P2)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN MAX	UNIT
t <sub>(ir</sub>	External interrupt timing	Port P1, P2: P1.x to P2.x, External trigger pulse width to set interrupt flag <sup>(1)</sup>	2.2 V/3 V	20	ns

<sup>(1)</sup> An external signal sets the interrupt flag every time the minimum interrupt pulse width t<sub>(int)</sub> is met. It may be set with trigger signals shorter than t<sub>(int)</sub>.

### Leakage Current (Ports P1, P2)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN MAX	UNIT
I <sub>lkg(Px.y)</sub>	High-impedance leakage current	(1) (2)	2.2 V/3 V	±50	nA

<sup>(1)</sup> The leakage current is measured with V<sub>SS</sub> or V<sub>CC</sub> applied to the corresponding pin(s), unless otherwise noted.

<sup>(2)</sup> The leakage of the digital port pins is measured individually. The port pin is selected for input and the pullup/pulldown resistor is disabled.



### **Outputs (Ports P1, P2)**

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	MAX	UNIT
V High level and a dealers	$I_{OH(max)} = -1.5 \text{ mA}^{(1)}$	2.2 V	V <sub>CC</sub> - 0.25	$V_{CC}$		
	High-level output voltage	$I_{OH(max)} = -6 \text{ mA}^{(2)}$	2.2 V	V <sub>CC</sub> - 0.6	$V_{CC}$	V
V <sub>OH</sub>	High-level output voltage	$I_{OH(max)} = -1.5 \text{ mA}^{(1)}$	2.1/	V <sub>CC</sub> - 0.25	$V_{CC}$	V
		$I_{OH(max)} = -6 \text{ mA}^{(2)}$	3 V	V <sub>CC</sub> - 0.6	$V_{CC}$	
		$I_{OL(max)} = 1.5 \text{ mA}^{(1)}$	2.2 V	V <sub>SS</sub>	$V_{SS} + 0.25$	
V	Low lovel output valtage	$I_{OL(max)} = 6 \text{ mA}^{(2)}$	2.2 V	V <sub>SS</sub>	$V_{SS} + 0.6$	V
V <sub>OL</sub>	Low-level output voltage	$I_{OL(max)} = 1.5 \text{ mA}^{(1)}$	3 V	V <sub>SS</sub>	$V_{SS} + 0.25$	V
		$I_{OL(max)} = 6 \text{ mA}^{(2)}$	3 V	V <sub>SS</sub>	V <sub>SS</sub> + 0.6	

<sup>(1)</sup> The maximum total current, I<sub>OH(max)</sub> and I<sub>OL(max)</sub>, for all outputs combined, should not exceed ±12 mA to hold the maximum voltage drop specified.

### **Output Frequency (Ports P1, P2)**

PARAMETER		TEST CONDITIONS	V <sub>CC</sub>	MIN MAX	UNIT
	Dort output fragues of (with load)	D4 4/SMCLK C 20 pF B 4 kQ(1)(2)	2.2 V	10	NAL 1-
T <sub>Px.y</sub>	Port output frequency (with load) P1.4/SMCLK, $C_L = 20 \text{ pF}$ , $R_L = 1 \text{ k}\Omega^{(1)(2)}$		3 V	12	MHz
	Clock cutout fraguency	D2 0/ACLK D4 4/SMCLK C 20 xE(2)	2.2 V	12	N/I I =
TPort_CLK	Clock output frequency	P2.0/ACLK, P1.4/SMCLK, $C_L = 20 \text{ pF}^{(2)}$	3 V	16	MHz

Alternatively, a resistive divider with two 0.5-kΩ resistors between V<sub>CC</sub> and V<sub>SS</sub> is used as load. The output is connected to the center tap of the divider.

<sup>(2)</sup> The maximum total current, I<sub>OH(max)</sub> and I<sub>OL(max)</sub>, for all outputs combined, should not exceed ±48 mA to hold the maximum voltage drop specified.

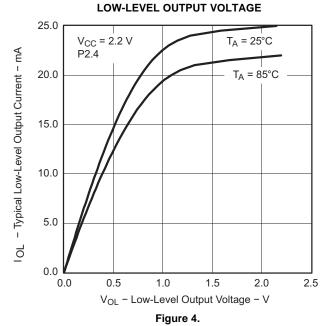
<sup>(2)</sup> The output voltage reaches at least 10% and 90% V<sub>CC</sub> at the specified toggle frequency.



### **Typical Characteristics - Outputs**

One output loaded at a time.

# TYPICAL LOW-LEVEL OUTPUT CURRENT vs



# OL - Typical Low-Level Output Current - mA

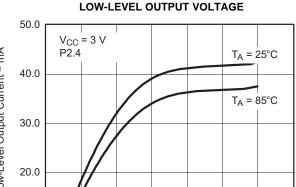
10.0

0.0

0.0

0.5

1.0



TYPICAL LOW-LEVEL OUTPUT CURRENT

V<sub>OL</sub> – Low-Level Output Voltage – V **Figure 5.** 

2.0

2.5

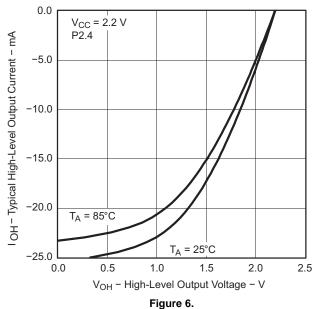
3.0

3.5

1.5

# TYPICAL HIGH-LEVEL OUTPUT CURRENT vs

### HIGH-LEVEL OUTPUT VOLTAGE



# TYPICAL HIGH-LEVEL OUTPUT CURRENT vs

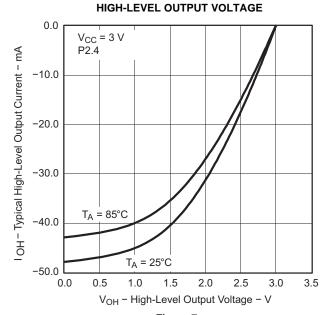


Figure 7.



### POR/Brownout Reset (BOR)<sup>(1)(2)</sup>

	PARAMETER	TEST CONDITIONS	T <sub>A</sub>	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
V <sub>CC(start)</sub>	See Figure 8	dV <sub>CC</sub> /dt ≤ 3 V/s				$0.7 \times V_{(B\_IT-)}$		٧
V <sub>(B_IT-)</sub>	See Figure 8 through Figure 10	dV <sub>CC</sub> /dt ≤ 3 V/s					1.71	٧
.,	On a Figure 0	111 / 11 4 0 1 / 1-	-40°C to 85°C		70	130	180	>/
V <sub>hys(B_IT-)</sub>	See Figure 8	dV <sub>CC</sub> /dt ≤ 3 V/s	105°C		70	130	210	mV
t <sub>d(BOR)</sub>	See Figure 8						2000	μs
t <sub>(reset)</sub>	Pulse length needed at RST/NMI pin to accepted reset internally			2.2 V/3 V	2			μs

- The current consumption of the brownout module is already included in the I<sub>CC</sub> current consumption data. The voltage level
- $V_{(B\_IT-)} + V_{hys(B\_IT-)}$  is  $\leq 1.8 \text{ V}$ . During power up, the CPU begins code execution following a period of  $t_{d(BOR)}$  after  $V_{CC} = V_{(B\_IT-)} + V_{hys(B\_IT-)}$ . The default DCO settings must not be changed until  $V_{CC} \geq V_{CC(min)}$ , where  $V_{CC(min)}$  is the minimum supply voltage for the desired operating frequency.

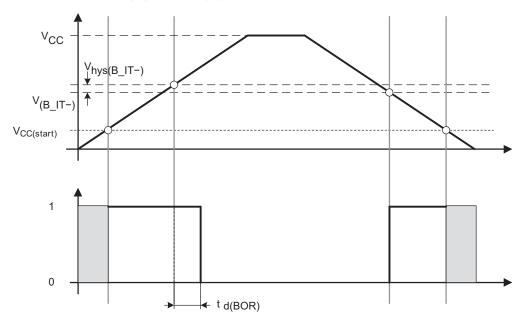


Figure 8. POR/Brownout Reset (BOR) vs Supply Voltage



### Typical Characteristics - POR/Brownout Reset (BOR)

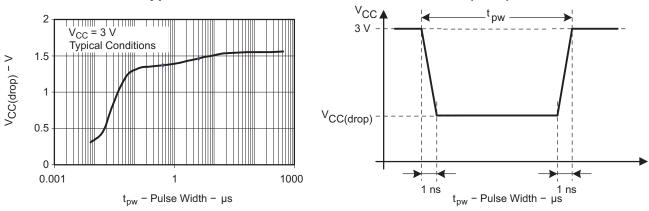


Figure 9.  $V_{\text{CC(drop)}}$  Level With a Square Voltage Drop to Generate a POR/Brownout Signal

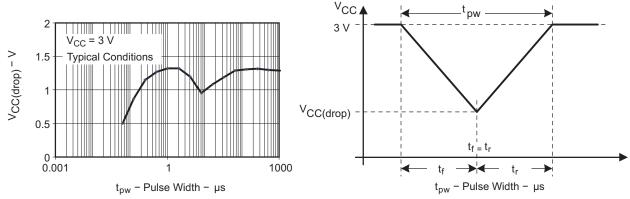


Figure 10. V<sub>CC(drop)</sub> Level With a Triangle Voltage Drop to Generate a POR/Brownout Signal



### **Main DCO Characteristics**

- All ranges selected by RSELx overlap with RSELx + 1: RSELx = 0 overlaps RSELx = 1, ... RSELx = 14 overlaps RSELx = 15.
- DCO control bits DCOx have a step size as defined by parameter S<sub>DCO</sub>.
- Modulation control bits MODx select how often f<sub>DCO(RSEL,DCO+1)</sub> is used within the period of 32 DCOCLK cycles. The frequency f<sub>DCO(RSEL,DCO)</sub> is used for the remaining cycles. The frequency is an average equal to:

$$f_{average} = \frac{32 \times f_{DCO(RSEL,DCO)} \times f_{DCO(RSEL,DCO+1)}}{MOD \times f_{DCO(RSEL,DCO)} + (32 - MOD) \times f_{DCO(RSEL,DCO+1)}}$$

### **DCO Frequency**

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
		RSELx < 14		1.8		3.6	
$V_{CC}$	Supply voltage range	RSELx = 14		2.2		3.6	V
		RSELx = 15		3.0		3.6	•
f <sub>DCO(0,0)</sub>	DCO frequency (0, 0)	RSELx = 0, $DCOx = 0$ , $MODx = 0$	2.2 V/3 V	0.06		0.14	MHz
f <sub>DCO(0,3)</sub>	DCO frequency (0, 3)	RSELx = 0, $DCOx = 3$ , $MODx = 0$	2.2 V/3 V	0.07		0.17	MHz
f <sub>DCO(1,3)</sub>	DCO frequency (1, 3)	RSELx = 1, DCOx = 3, MODx = 0	2.2 V/3 V	0.10		0.20	MHz
f <sub>DCO(2,3)</sub>	DCO frequency (2, 3)	RSELx = 2, $DCOx = 3$ , $MODx = 0$	2.2 V/3 V	0.14		0.28	MHz
f <sub>DCO(3,3)</sub>	DCO frequency (3, 3)	RSELx = 3, $DCOx = 3$ , $MODx = 0$	2.2 V/3 V	0.20		0.40	MHz
f <sub>DCO(4,3)</sub>	DCO frequency (4, 3)	RSELx = 4, DCOx = 3, MODx = 0	2.2 V/3 V	0.28		0.54	MHz
f <sub>DCO(5,3)</sub>	DCO frequency (5, 3)	RSELx = 5, $DCOx = 3$ , $MODx = 0$	2.2 V/3 V	0.39		0.77	MHz
f <sub>DCO(6,3)</sub>	DCO frequency (6, 3)	RSELx = 6, $DCOx = 3$ , $MODx = 0$	2.2 V/3 V	0.54		1.06	MHz
f <sub>DCO(7,3)</sub>	DCO frequency (7, 3)	RSELx = 7, DCOx = 3, MODx = 0	2.2 V/3 V	0.80		1.50	MHz
f <sub>DCO(8,3)</sub>	DCO frequency (8, 3)	RSELx = 8, $DCOx = 3$ , $MODx = 0$	2.2 V/3 V	1.10		2.10	MHz
f <sub>DCO(9,3)</sub>	DCO frequency (9, 3)	RSELx = 9, $DCOx = 3$ , $MODx = 0$	2.2 V/3 V	1.60		3.00	MHz
f <sub>DCO(10,3)</sub>	DCO frequency (10, 3)	RSELx = 10, DCOx = 3, MODx = 0	2.2 V/3 V	2.50		4.30	MHz
f <sub>DCO(11,3)</sub>	DCO frequency (11, 3)	RSELx = 11, DCOx = 3, MODx = 0	2.2 V/3 V	3.00		5.50	MHz
f <sub>DCO(12,3)</sub>	DCO frequency (12, 3)	RSELx = 12, DCOx = 3, MODx = 0	2.2 V/3 V	4.30		7.30	MHz
f <sub>DCO(13,3)</sub>	DCO frequency (13, 3)	RSELx = 13, DCOx = 3, MODx = 0	2.2 V/3 V	6.00		9.60	MHz
f <sub>DCO(14,3)</sub>	DCO frequency (14, 3)	RSELx = 14, DCOx = 3, MODx = 0	2.2 V/3 V	8.60		13.9	MHz
f <sub>DCO(15,3)</sub>	DCO frequency (15, 3)	RSELx = 15, DCOx = 3, MODx = 0	3 V	12.0		18.5	MHz
f <sub>DCO(15,7)</sub>	DCO frequency (15, 7)	RSELx = 15, DCOx = 7, MODx = 0	3 V	16.0		26.0	MHz
S <sub>RSEL</sub>	Frequency step between range RSEL and RSEL+1	$S_{RSEL} = f_{DCO(RSEL+1,DCO)} / f_{DCO(RSEL,DCO)}$	2.2 V/3 V			1.55	ratio
S <sub>DCO</sub>	Frequency step between tap DCO and DCO+1	$S_{DCO} = f_{DCO(RSEL,DCO+1)} / f_{DCO(RSEL,DCO)}$	2.2 V/3 V	1.05	1.08	1.12	ratio
	Duty cycle	Measured at P1.4/SMCLK	2.2 V/3 V	40	50	60	%



### **Calibrated DCO Frequencies - Tolerance at Calibration**

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	T <sub>A</sub>	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
	Frequency tolerance at calibration		25°C	3 V	-1	±0.2	+1	%
f <sub>CAL(1MHz)</sub>	1-MHz calibration value	BCSCTL1 = CALBC1_1MHZ, DCOCTL = CALDCO_1MHZ, Gating time: 5 ms	25°C	3 V	0.990	1	1.010	MHz
f <sub>CAL(8MHz)</sub>	8-MHz calibration value	BCSCTL1 = CALBC1_8MHZ, DCOCTL = CALDCO_8MHZ, Gating time: 5 ms	25°C	3 V	7.920	8	8.080	MHz
f <sub>CAL(12MHz)</sub>	12-MHz calibration value	BCSCTL1 = CALBC1_12MHZ, DCOCTL = CALDCO_12MHZ, Gating time: 5 ms	25°C	3 V	11.88	12	12.12	MHz
f <sub>CAL(16MHz)</sub>	16-MHz calibration value	BCSCTL1 = CALBC1_16MHZ, DCOCTL = CALDCO_16MHZ, Gating time: 2 ms	25°C	3 V	15.84	16	16.16	MHz

### Calibrated DCO Frequencies - Tolerance Over Temperature 0°C to 85°C

	PARAMETER	TEST CONDITIONS	T <sub>A</sub>	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
	1-MHz tolerance over temperature		0°C to 85°C	3 V	-2.5	±0.5	+2.5	%
	8-MHz tolerance over temperature		0°C to 85°C	3 V	-2.5	±1	+2.5	%
	12-MHz tolerance over temperature		0°C to 85°C	3 V	-2.5	±1	+2.5	%
	16-MHz tolerance over temperature		0°C to 85°C	3 V	-3	±2	+3	%
		BCSCTL1 = CALBC1_1MHZ,		2.2 V	0.97	1	1.03	
f <sub>CAL(1MHz)</sub>	1-MHz calibration value	DCOCTL = CALDCO_1MHZ,	0°C to 85°C	3 V	0.975	1	1.025	MHz
		Gating time: 5 ms		3.6 V	0.97	1	1.03	
		BCSCTL1 = CALBC1_8MHZ, DCOCTL = CALDCO_8MHZ,		2.2 V	7.76	8	8.4	
f <sub>CAL(8MHz)</sub>			0°C to 85°C	3 V	7.8	8	8.2	MHz
		Gating time: 5 ms		3.6 V	7.6	8	8.24	·
		BCSCTL1 = CALBC1 12MHZ,		2.2 V	11.7	12	12.3	
f <sub>CAL(12MHz)</sub>	12-MHz calibration value	DCOCTL = CALDCO_12MHZ,	0°C to 85°C	3 V	11.7	12	12.3	MHz
		Gating time: 5 ms		3.6 V	11.7	12	12.3	·
		BCSCTL1 = CALBC1_16MHZ,		3 V	15.52	16	16.48	
f <sub>CAL(16MHz)</sub>	16-MHz calibration value	DCOCTL = CALDCO_16MHZ, Gating time: 2 ms	0°C to 85°C	3.6 V	15	16	16.48	MHz



## Calibrated DCO Frequencies - Tolerance Over Supply Voltage $V_{\text{CC}}$

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	T <sub>A</sub>	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
	1-MHz tolerance over V <sub>CC</sub>		25°C	1.8 V to 3.6 V	-3	±2	+3	%
	8-MHz tolerance over V <sub>CC</sub>		25°C	1.8 V to 3.6 V	-3	±2	+3	%
	12-MHz tolerance over $V_{CC}$		25°C	2.2 V to 3.6 V	-3	±2	+3	%
	16-MHz tolerance over V <sub>CC</sub>		25°C	3 V to 3.6 V	-3	±2	+3	%
f <sub>CAL(1MHz)</sub>	1-MHz calibration value	BCSCTL1 = CALBC1_1MHZ, DCOCTL = CALDCO_1MHZ, Gating time: 5 ms	25°C	1.8 V to 3.6 V	0.97	1	1.03	MHz
f <sub>CAL(8MHz)</sub>	8-MHz calibration value	BCSCTL1 = CALBC1_8MHZ, DCOCTL = CALDCO_8MHZ, Gating time: 5 ms	25°C	1.8 V to 3.6 V	7.76	8	8.24	MHz
f <sub>CAL(12MHz)</sub>	12-MHz calibration value	BCSCTL1 = CALBC1_12MHZ, DCOCTL = CALDCO_12MHZ, Gating time: 5 ms	25°C	2.2 V to 3.6 V	11.64	12	12.36	MHz
f <sub>CAL(16MHz)</sub>	16-MHz calibration value	BCSCTL1 = CALBC1_16MHZ, DCOCTL = CALDCO_16MHZ, Gating time: 2 ms	25°C	3 V to 3.6 V	15	16	16.48	MHz

### **Calibrated DCO Frequencies - Overall Tolerance**

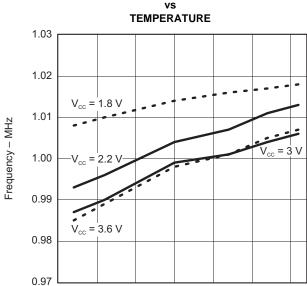
PAF	RAMETER	TEST CONDITIONS	T <sub>A</sub>	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
	1-MHz tolerance overall		I: -40°C to 85°C T: -40°C to 105°C	1.8 V to 3.6 V	-5	±2	+5	%
	8-MHz tolerance overall		I: -40°C to 85°C T: -40°C to 105°C	1.8 V to 3.6 V	-5	±2	+5	%
	12-MHz tolerance overall		I: -40°C to 85°C T: -40°C to 105°C	2.2 V to 3.6 V	-5	±2	+5	%
	16-MHz tolerance overall		I: -40°C to 85°C T: -40°C to 105°C	3 V to 3.6 V	-6	±3	+6	%
f <sub>CAL(1MHz)</sub>	1-MHz calibration value	BCSCTL1 = CALBC1_1MHZ, DCOCTL = CALDCO_1MHZ, Gating time: 5 ms	I: -40°C to 85°C T: -40°C to 105°C	1.8 V to 3.6 V	0.95	1	1.05	MHz
f <sub>CAL(8MHz)</sub>	8-MHz calibration value	BCSCTL1 = CALBC1_8MHZ, DCOCTL = CALDCO_8MHZ, Gating time: 5 ms	I: -40°C to 85°C T: -40°C to 105°C	1.8 V to 3.6 V	7.6	8	8.4	MHz
f <sub>CAL(12MHz)</sub>	12-MHz calibration value	BCSCTL1 = CALBC1_12MHZ, DCOCTL = CALDCO_12MHZ, Gating time: 5 ms	I: -40°C to 85°C T: -40°C to 105°C	2.2 V to 3.6 V	11.4	12	12.6	MHz
f <sub>CAL(16MHz)</sub>	16-MHz calibration value	BCSCTL1 = CALBC1_16MHZ, DCOCTL = CALDCO_16MHZ, Gating time: 2 ms	I: -40°C to 85°C T: -40°C to 105°C	3 V to 3.6 V	15	16	17	MHz



-50

-25

### Typical Characteristics - Calibrated 1-MHz DCO Frequency **CALIBRATED 1-MHz FREQUENCY**



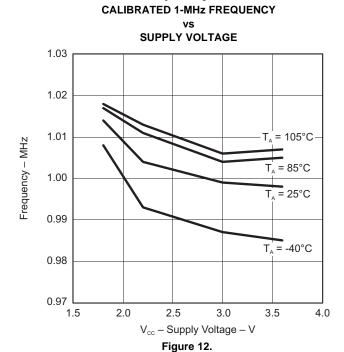
25

Figure 11.

T<sub>A</sub> - Temperature - °C

# 50 75

100





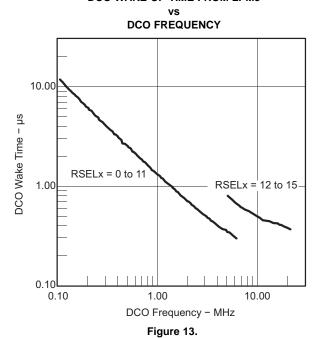
### Wake-Up From Lower-Power Modes (LPM3/4)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
		BCSCTL1 = CALBC1_1MHZ, DCOCTL = CALDCO_1MHZ				2	
	DCO clock wake-up time	BCSCTL1 = CALBC1_8MHZ, DCOCTL = CALDCO_8MHZ	2.2 V/3 V			1.5	
<sup>T</sup> DCO,LPM3/4	from LPM3/4 <sup>(1)</sup>	BCSCTL1 = CALBC1_12MHZ, DCOCTL = CALDCO_12MHZ				1	μs
		BCSCTL1 = CALBC1_16MHZ, DCOCTL = CALDCO_16MHZ	3 V			1	
t <sub>CPU,LPM3/4</sub>	CPU wake-up time from LPM3/4 <sup>(2)</sup>				1 / f <sub>MCLK</sub> + t <sub>Clock,LPM3/4</sub>		

<sup>(1)</sup> The DCO clock wake-up time is measured from the edge of an external wake-up signal (for example, a port interrupt) to the first clock edge observable externally on a clock pin (MCLK or SMCLK).

# Typical Characteristics - DCO Clock Wake-Up Time From LPM3/4 DCO WAKE-UP TIME FROM LPM3



<sup>(2)</sup> Parameter applicable only if DCOCLK is used for MCLK.



### Crystal Oscillator LFXT1, Low-Frequency Mode<sup>(1)</sup>

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
f <sub>LFXT1,LF</sub>	LFXT1 oscillator crystal frequency, LF mode 0, 1	XTS = 0, LFXT1Sx = 0 or 1	1.8 V to 3.6 V		32768		Hz
f <sub>LFXT1,LF,logic</sub>	LFXT1 oscillator logic level square wave input frequency, LF mode	XTS = 0, LFXT1Sx = 3	1.8 V to 3.6 V	10000	32768	50000	Hz
04	Oscillation allowance for	$XTS = 0$ , $LFXT1Sx = 0$ , $f_{LFXT1,LF} = 32768$ Hz, $C_{L,eff} = 6$ pF			500		kΩ
OA <sub>LF</sub>	LF crystals	$XTS = 0$ , $LFXT1Sx = 0$ , $f_{LFXT1,LF} = 32768$ Hz, $C_{L,eff} = 12$ pF			200		K12
		XTS = 0, $XCAPx = 0$			1		
<u> </u>	Integrated effective load	XTS = 0, $XCAPx = 1$			5.5		~F
$C_{L,eff}$	capacitance, LF mode (2)	XTS = 0, $XCAPx = 2$			8.5		pF
		XTS = 0, XCAPx = 3			11		
	Duty cycle, LF mode	XTS = 0, Measured at P2.0/ACLK, f <sub>LFXT1,LF</sub> = 32768 Hz	2.2 V/3 V	30	50	70	%
f <sub>Fault,LF</sub>	Oscillator fault frequency, LF mode <sup>(3)</sup>	XTS = 0, LFXT1Sx = 3 <sup>(4)</sup>	2.2 V/3 V	10		10000	Hz

- (1) To improve EMI on the XT1 oscillator, the following guidelines should be observed.
  - (a) Keep the trace between the device and the crystal as short as possible.
  - (b) Design a good ground plane around the oscillator pins.
  - (c) Prevent crosstalk from other clock or data lines into oscillator pins XIN and XOUT.
  - (d) Avoid running PCB traces underneath or adjacent to the XIN and XOUT pins.
  - (e) Use assembly materials and praxis to avoid any parasitic load on the oscillator XIN and XOUT pins.
  - (f) If conformal coating is used, ensure that it does not induce capacitive/resistive leakage between the oscillator pins.
  - (g) Do not route the XOUT line to the JTAG header to support the serial programming adapter as shown in other documentation. This signal is no longer required for the serial programming adapter.
- (2) Includes parasitic bond and package capacitance (approximately 2 pF per pin).
  - Because the PCB adds additional capacitance, it is recommended to verify the correct load by measuring the ACLK frequency. For a correct setup, the effective load capacitance should always match the specification of the crystal that is used.
- (3) Frequencies below the MIN specification set the fault flag. Frequencies above the MAX specification do not set the fault flag. Frequencies in between might set the flag.
- (4) Measured with logic-level input frequency but also applies to operation with crystals.



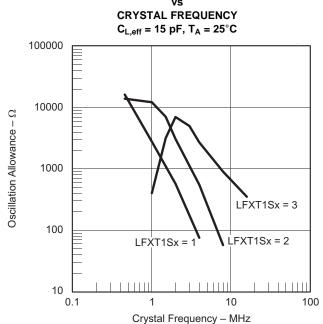
### Crystal Oscillator LFXT1, High-Frequency Mode<sup>(1)</sup>

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
f <sub>LFXT1,HF0</sub>	LFXT1 oscillator crystal frequency, HF mode 0	XTS = 1, LFXT1Sx = 0	1.8 V to 3.6 V	0.4		1	MHz
f <sub>LFXT1,HF1</sub>	LFXT1 oscillator crystal frequency, HF mode 1	XTS = 1, LFXT1Sx = 1	1.8 V to 3.6 V	1		4	MHz
			1.8 V to 3.6 V	2		10	
f <sub>LFXT1,HF2</sub>	LFXT1 oscillator crystal frequency, HF mode 2	XTS = 1, LFXT1Sx = 2	2.2 V to 3.6 V	2		12	MHz
	nequency, in mode 2		3 V to 3.6 V	2		16	
	LFXT1 oscillator logic-level		1.8 V to 3.6 V	0.4		10	
f <sub>LFXT1,HF,logic</sub>	square-wave input	XTS = 1, LFXT1Sx = 3	2.2 V to 3.6 V	0.4		12	MHz
	frequency, HF mode		3 V to 3.6 V	0.4		16	
		$XTS = 1$ , $LFXT1Sx = 0$ , $f_{LFXT1,HF} = 1$ MHz, $C_{L,eff} = 15$ pF			2700		
OA <sub>HF</sub>	Oscillation allowance for HF crystals (see Figure 14 and Figure 15)	$XTS = 1$ , $LFXT1Sx = 1$ , $f_{LFXT1,HF} = 4$ MHz, $C_{L,eff} = 15$ pF			800		Ω
	riguio 10)	$XTS = 1$ , $LFXT1Sx = 2$ , $f_{LFXT1,HF} = 16$ MHz, $C_{L,eff} = 15$ pF			300		
$C_{L,eff}$	Integrated effective load capacitance, HF mode <sup>(2)</sup>	XTS = 1 <sup>(3)</sup>			1		pF
	Duty avalo LIE made	XTS = 1, Measured at P2.0/ACLK, f <sub>LFXT1,HF</sub> = 10 MHz	2 2 3 7 7	40	50	60	0/
	Duty cycle, HF mode	XTS = 1, Measured at P2.0/ACLK, f <sub>LFXT1,HF</sub> = 16 MHz	2.2 V/3 V	40	50	60	%
f <sub>Fault,HF</sub>	Oscillator fault frequency (4)	XTS = 1, LFXT1Sx = 3 <sup>(5)</sup>	2.2 V/3 V	30		300	kHz

- (1) To improve EMI on the XT2 oscillator the following guidelines should be observed:
  - (a) Keep the trace between the device and the crystal as short as possible.
  - (b) Design a good ground plane around the oscillator pins.
  - (c) Prevent crosstalk from other clock or data lines into oscillator pins XIN and XOUT.
  - (d) Avoid running PCB traces underneath or adjacent to the XIN and XOUT pins.
  - (e) Use assembly materials and praxis to avoid any parasitic load on the oscillator XIN and XOUT pins.
  - (f) If conformal coating is used, ensure that it does not induce capacitive/resistive leakage between the oscillator pins.
  - (g) Do not route the XOUT line to the JTAG header to support the serial programming adapter as shown in other documentation. This signal is no longer required for the serial programming adapter.
- (2) Includes parasitic bond and package capacitance (approximately 2 pF per pin). Because the PCB adds additional capacitance, it is recommended to verify the correct load by measuring the ACLK frequency. For a correct setup, the effective load capacitance should always match the specification of the used crystal.
- (3) Requires external capacitors at both terminals. Values are specified by crystal manufacturers.
- (4) Frequencies below the MIN specification set the fault flag, frequencies above the MAX specification do not set the fault flag, and frequencies in between might set the flag.
- (5) Measured with logic-level input frequency, but also applies to operation with crystals.



# Typical Characteristics - LFXT1 Oscillator in HF Mode (XTS = 1) OSCILLATION ALLOWANCE OSCILLATOR SUPPLY CURRENT



# Figure 14.

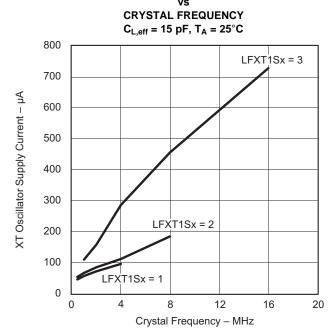


Figure 15.



### Timer\_A

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP MAX	UNIT
		Internal: SMCLK, ACLK	2.2 V		10	
f <sub>TA</sub>	Timer_A clock frequency	External: TACLK, INCLK Duty cycle = 50% ± 10%	3 V		16	MHz
t <sub>TA,cap</sub>	Timer_A capture timing	TA0, TA1, TA2	2.2 V/3 V	20		ns

# Comparator\_A+(1)

over recommended operating free-air temperature range (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
		CAON 4 CARCEL A CAREE A	2.2 V		25	40	
I <sub>(DD)</sub>		CAON = 1, CARSEL = 0, CAREF = 0	3 V		45	60	μA
		CAON = 1, CARSEL = 0, CAREF = 1/2/3,	2.2 V		30	50	
(Refladder/R	efDiode)	No load at P2.3/CA0/TA1 and P2.4/CA1/TA2	3 V		45	71	μA
V <sub>(IC)</sub>	Common-mode input voltage range	CAON = 1	2.2 V/3 V	0		V <sub>CC</sub> - 1	V
V <sub>(Ref025)</sub>	(Voltage at 0.25 V <sub>CC</sub> node) / V <sub>CC</sub>	PCA0 = 1, CARSEL = 1, CAREF = 1, No load at P2.3/CA0/TA1 and P2.4/CA1/TA2	2.2 V/3 V	0.23	0.24	0.25	
V <sub>(Ref050)</sub>	(Voltage at 0.5 V <sub>CC</sub> node) / V <sub>CC</sub>	PCA0 = 1, CARSEL = 1, CAREF = 2, No load at P2.3/CA0/TA1 and P2.4/CA1/TA2	2.2 V/3 V	0.47	0.48	0.5	
	See Figure 19 and	PCA0 = 1, CARSEL = 1, CAREF = 3,	2.2 V	390	480	540	
V <sub>(RefVT)</sub>	Figure 20	No load at P2.3/CA0/TA1 and P2.4/CA1/TA2, $T_A = 85^{\circ}C$	3 V	400	490	550	mV
V <sub>(offset)</sub>	Offset voltage (2)		2.2 V/3 V	-30		30	mV
$V_{\text{hys}}$	Input hysteresis	CAON = 1	2.2 V/3 V	0	0.7	1.4	mV
		$T_A = 25^{\circ}C$ , Overdrive 10 mV,	2.2 V	80	165	300	
	Response time	Without filter: CAF = 0 <sup>(3)</sup> (see Figure 16 and Figure 17)	3 V	70	120	240	ns
t <sub>(response)</sub>	(low-high and high-low)	$T_A = 25^{\circ}C$ , Overdrive 10 mV,	2.2 V	1.4	1.9	2.8	
		With filter: CAF = 1 <sup>(3)</sup> (see Figure 16 and Figure 17)	3 V	0.9	1.5	2.2	μs

 <sup>(1)</sup> The leakage current for the Comparator\_A+ terminals is identical to I<sub>lkg(Px,y)</sub> specification.
 (2) The input offset voltage can be cancelled by using the CAEX bit to invert the Comparator\_A+ inputs on successive measurements. The two successive measurements are then summed together.

Response time measured at P2.2/CAOUT.



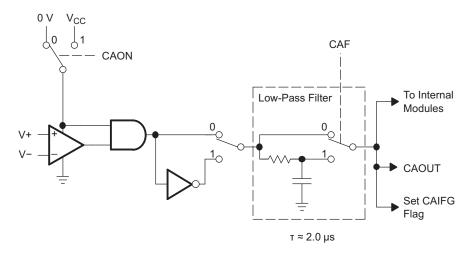


Figure 16. Comparator\_A+ Module Block Diagram

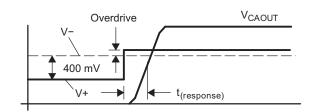


Figure 17. Overdrive Definition

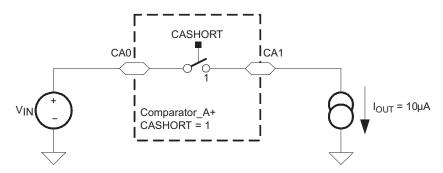
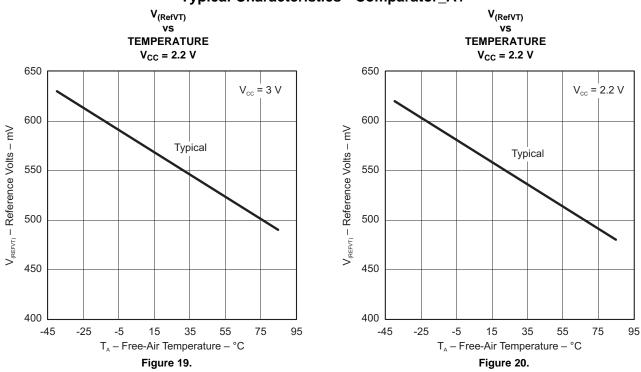


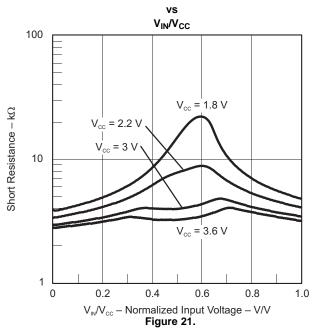
Figure 18. Comparator\_A+ Short Resistance Test Condition



### Typical Characteristics - Comparator\_A+



### SHORT RESISTANCE





### **Flash Memory**

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
V <sub>CC</sub> (PGM/ERASE)	Program and erase supply voltage			2.2		3.6	V
f <sub>FTG</sub>	Flash timing generator frequency			257		476	kHz
I <sub>PGM</sub>	Supply current from V <sub>CC</sub> during program		2.2 V/3.6 V		3	5	mA
I <sub>ERASE</sub>	Supply current from V <sub>CC</sub> during erase		2.2 V/3.6 V		3	7	mA
t <sub>CPT</sub>	Cumulative program time <sup>(1)</sup>		2.2 V/3.6 V			10	ms
t <sub>CMErase</sub>	Cumulative mass erase time		2.2 V/3.6 V	20			ms
	Program/erase endurance			10 <sup>4</sup>	10 <sup>5</sup>		cycles
t <sub>Retention</sub>	Data retention duration	$T_J = 25^{\circ}C$		100			years
t <sub>Word</sub>	Word or byte program time	See (2)			30		t <sub>FTG</sub>
t <sub>Block, 0</sub>	Block program time for first byte or word	See (2)			25		t <sub>FTG</sub>
t <sub>Block, 1-63</sub>	Block program time for each additional byte or word	See (2)			18		t <sub>FTG</sub>
t <sub>Block, End</sub>	Block program end-sequence wait time	See (2)			6		t <sub>FTG</sub>
t <sub>Mass Erase</sub>	Mass erase time	See (2)			10593		t <sub>FTG</sub>
t <sub>Seg Erase</sub>	Segment erase time	See (2)			4819		t <sub>FTG</sub>

<sup>(1)</sup> The cumulative program time must not be exceeded when writing to a 64-byte flash block. This parameter applies to all programming methods: individual word/byte write and block write modes.

### **RAM**

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN MAX	UNIT
$V_{(RAMh)}$	RAM retention supply voltage <sup>(1)</sup>	CPU halted	1.6	V

<sup>(1)</sup> This parameter defines the minimum supply voltage V<sub>CC</sub> when the data in RAM remains unchanged. No program execution should happen during this supply voltage condition.

### **JTAG Interface**

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

		`		,		
	PARAMETER	V <sub>CC</sub>	MIN	TYP	MAX	TINU
	TCK input frequency <sup>(1)</sup>	2.2 V	0		5	MHz
ITCK	TCK input frequency (7)	3 V	0		10	MHz
R <sub>Internal</sub>	Internal pulldown resistance on TEST	2.2 V/3 V	25	60	90	kΩ

<sup>(1)</sup> f<sub>TCK</sub> may be restricted to meet the timing requirements of the module selected.

### JTAG Fuse<sup>(1)</sup>

	PARAMETER				UNIT
V <sub>CC(FB)</sub>	Supply voltage during fuse-blow condition	25°C	2.5		V
$V_{FB}$	Voltage level on TEST for fuse blow	25°C	6	7	V
I <sub>FB</sub>	Supply current into TEST during fuse blow	25°C		100	mA
t <sub>FB</sub>	Time to blow fuse	25°C		1	ms

<sup>(1)</sup> Once the fuse is blown, no further access to the JTAG/Test and emulation features is possible, and the JTAG block is switched to bypass mode.

<sup>2)</sup> These values are hardwired into the flash controller's state machine ( $t_{FTG} = 1/f_{FTG}$ ).



### **APPLICATION INFORMATION**

### Port P1 Pin Schematic: P1.0 to P1.3, Input/Output With Schmitt Trigger

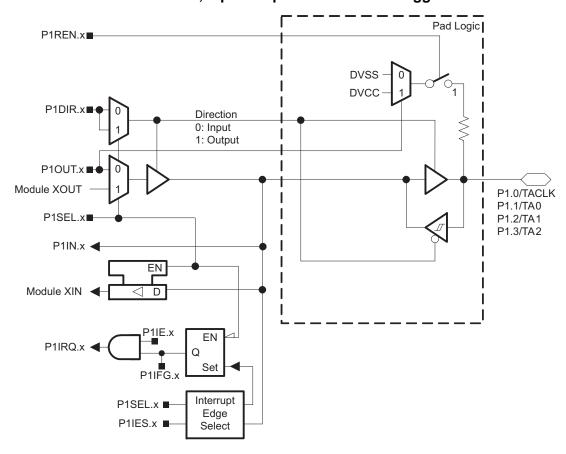


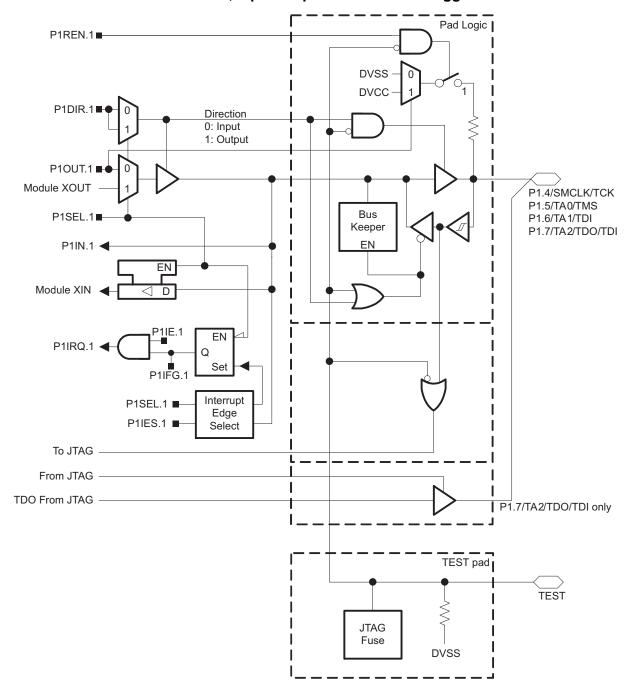
Table 17. Port P1 (P1.0 to P1.3) Pin Functions

DINI NIAME (D4)	х	FUNCTION	CONTROL BITS / SIGNALS	
PIN NAME (P1.x)		FUNCTION	P1DIR.x	P1SEL.x
		P1.0 <sup>(1)</sup> (I/O)	I: 0; O: 1	0
P1.0/TACLK	0	TACLK	0	1
		DVSS	1	1
		P1.1 <sup>(1)</sup> (I/O)	I: 0; O: 1	0
P1.1/TA0	1	Timer_A3.CCI0A	0	1
		Timer_A3.TA0	1	1
		P1.2 <sup>(1)</sup> (I/O)	I: 0; O: 1	0
P1.2/TA1	2	Timer_A3.CCI0A	0	1
		Timer_A3.TA0	1	1
		P1.3 <sup>(1)</sup> (I/O)	I: 0; O: 1	0
P1.3/TA2	3	Timer_A3.CCI0A	0	1
		Timer_A3.TA0	1	1

<sup>(1)</sup> Default after reset (PUC/POR)



### Port P1 Pin Schematic: P1.4 to P1.7, Input/Output With Schmitt Trigger





### Table 18. Port P1 (P1.4 to P1.7) Pin Functions

DIN NAME (D4 v)		FUNCTION	CON	CONTROL BITS / SIGNALS <sup>(1)</sup>			
PIN NAME (P1.x)	X		P1DIR.x	P1SEL.x	TEST		
		P1.4 <sup>(2)</sup> (I/O)	I: 0; O: 1	0	0		
P1.4/SMCLK/TCK	4	SMCLK	1	1	0		
		TCK	X	Х	1		
		P1.5 <sup>(2)</sup> (I/O)	I: 0; O: 1	0	0		
P1.5/TA0/TMS	5	Timer_A3.TA0	1	1	0		
		TMS	X	Х	1		
		P1.6 <sup>(2)</sup> (I/O)	I: 0; O: 1	0	0		
P1.6/TA1/TDI/TCLK	6	Timer_A3.TA1	1	1	0		
		TDI/TCLK <sup>(3)</sup>	X	Х	1		
		P1.7 <sup>(2)</sup> (I/O)	I: 0; O: 1	0	0		
P1.7/TA2/TDO/TDI	7	Timer_A3.TA2	1	1	0		
		TDO/TDI <sup>(3)</sup>	Х	Х	1		

<sup>(1)</sup> X = don't care

 <sup>(2)</sup> Default after reset (PUC/POR)
 (3) Function controlled by JTAG



### Port P2 Pin Schematic: P2.0 to P2.5, Input/Output With Schmitt Trigger

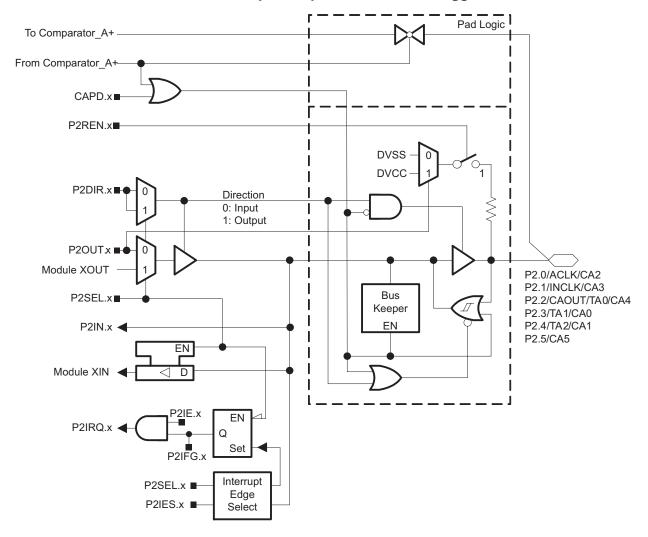


Table 19. Control Signal "From Comparator\_A+"

DINI NAME	FUNCTION	SIGNAL "From Comparator_A+" = 1 <sup>(1)</sup>									
PIN NAME	FUNCTION	P2CA4	P2CA0		P2CA3	P2CA2	P2CA1				
P2.0/ACLK/CA2	CA2	1	1		0	1	0				
P2.1/INCLK/CA3	CA3	N/A	N/A		0	1	1				
P2.2/CAOUT/TA0/CA4	CA4	N/A	N/A	<b>O</b> D	1	0	0				
P2.3/TA1/CA0	CA0	0	1	OR	N/A	N/A	N/A				
P2.4/TA2/CA1	CA1	1	0		0	0	1				
P2.5/CA5	CA5	N/A	N/A		1	0	1				

<sup>(1)</sup> N/A = Not available or not applicable



### Table 20. Port P2 (P2.0 to P2.5) Pin Functions

DIN NAME (DO. )		FUNCTION	CON	TROL BITS / SIGNA	\LS <sup>(1)</sup>
PIN NAME (P2.x)	X	FUNCTION	P2DIR.x	P2SEL.x	CAPD.x
		P2.0 <sup>(2)</sup> (I/O)	I: 0; O: 1	0	0
P2.0/ACLK/CA2	0	ACLK	1	1	0
		CA2 <sup>(3)</sup>	Х	Х	1
		P2.1 <sup>(2)</sup> (I/O)	I: 0; O: 1	0	0
20 4/INCLIV/CA2	1	Timer_A3.INCLK	0	1	0
P2.1/INCLK/CA3	1	DVSS	1	1	0
		CA3 <sup>(3)</sup>	Х	Х	1
		P2.2 <sup>(2)</sup> (I/O)	I: 0; O: 1	0	0
	2	Timer_A3.CCI0B	0	1	0
P2.2/CAOUT/TA0/CA4	2	CAOUT	1	1	0
		CA4 <sup>(3)</sup>	Х	Х	1
		P2.3 <sup>(2)</sup> (I/O)	I: 0; O: 1	0	0
P2.3/TA1/CA0	3	Timer_A3.TA1	1	1	0
		CA0 <sup>(3)</sup>	Х	Х	1
		P2.4 <sup>(2)</sup> (I/O)	I: 0; O: 1	0	0
P2.4/TA2/CA1	4	Timer_A3.TA2	1	1	0
		CA1 <sup>(3)</sup>	Х	Х	1
20.5/045	_	P2.5 <sup>(2)</sup> (I/O)	I: 0; O: 1	0	0
P2.5/CA5	5	CA5 <sup>(3)</sup>	X	Х	1

X = don't care

Default after reset (PUC/POR)

Setting theCAPD.x bit disables the output driver as well as the input Schmitt trigger to prevent parasitic cross currentswhen applying analog signals. Selecting the CAx input pin to the comparator multiplexer with the P2CAx bits automatically disables the input buffer for that pin, regardless of the state of the associated CAPD.x bit.



### Port P2 Pin Schematic: P2.6, Input/Output With Schmitt Trigger and Crystal Oscillator Input

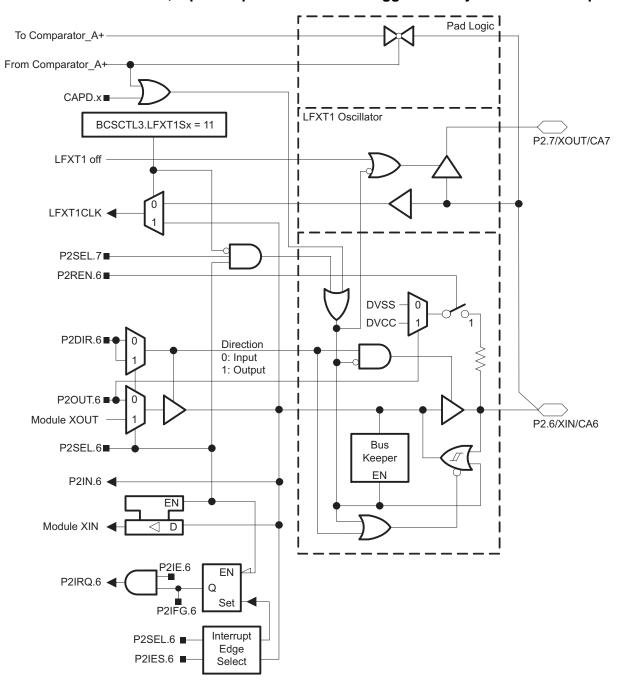


Table 21. Control Signal "From Comparator\_A+"

PIN NAME	FUNCTION	SIGNAL "From Comparator_A+" = 1						
PIN NAME	FUNCTION	P2CA3	P2CA2	P2CA1				
P2.6/XIN/CA6	CA6	1	1	0				



### Table 22. Port P2 (P2.6) Pin Functions

PIN NAME (P2.x)		FUNCTION	CONTROL BITS / SIGNALS <sup>(1)</sup>					
	Х	FUNCTION	P2DIR.x	P2SEL.x	CAPD.x			
P2.6/XIN/CA6		P2.6 (I/O)	I: 0; O: 1	0	0			
	6	XIN <sup>(2)</sup>	Х	1	0			
		CA6 <sup>(3)</sup>	Х	Х	1			

X = don't care

Default after reset (PUC/POR)
Setting theCAPD.x bit disables the output driver as well as the input Schmitt trigger to prevent parasitic cross currentswhen applying analog signals. Selecting the CAx input pin to the comparator multiplexer with the P2CAx bits automatically disables the input buffer for that pin, regardless of the state of the associated CAPD.x bit.



### Port P2 Pin Schematic: P2.7, Input/Output With Schmitt Trigger and Crystal Oscillator Output

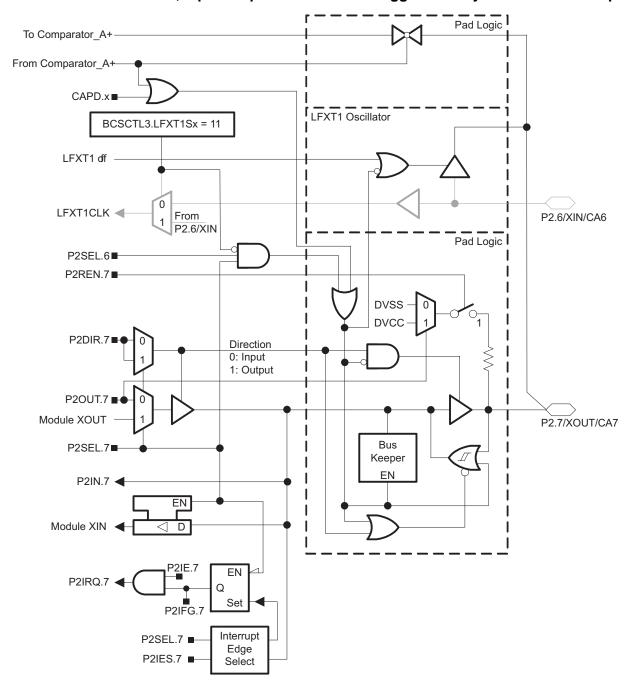


Table 23. Control Signal "From Comparator\_A+"

PIN NAME	FUNCTION	SIGNAL "From Comparator_A+" = 1						
	FUNCTION	P2CA3	P2CA2	P2CA1				
P2.7/XOUT/CA7	CA7	1	1	1				



### Table 24. Port P2 (P2.7) Pin Functions

PIN NAME (P2.x)		FUNCTION	CONTROL BITS / SIGNALS <sup>(1)</sup>					
	X	FUNCTION	P2DIR.x	P2SEL.x	CAPD.x			
P2.7/XOUT/CA7		P2.7 (I/O)	I: 0; O: 1	0	0			
	6	XOUT <sup>(2)(3)</sup>	Х	1	0			
		CA7 <sup>(4)</sup>	Х	Х	1			

- (1) X = don't care
- (2) Default after reset (PUC/POR)
- (3) If the pin XOUT/P2.7/CA7 is used as an input a current can flow until P2SEL.7 is cleared due to the oscillator output driver connection to this pin after reset.
   (4) Setting the CAPD.x bit disables the output driver as well as the input Schmitt trigger to prevent parasitic cross currents when applying
- (4) Setting the CAPD.x bit disables the output driver as well as the input Schmitt trigger to prevent parasitic cross currentswhen applying analog signals. Selecting the CAx input pin to the comparator multiplexer with the P2CAx bits automatically disables the input buffer for that pin, regardless of the state of the associated CAPD.x bit.



#### **JTAG Fuse Check Mode**

MSP430 devices that have the fuse on the TEST terminal have a fuse check mode that tests the continuity of the fuse the first time the JTAG port is accessed after a power-on reset (POR). When activated, a fuse check current,  $I_{TF}$ , of 1 mA at 3 V, 2.5 mA at 5 V can flow from the TEST pin to ground if the fuse is not burned. Care must be taken to avoid accidentally activating the fuse check mode and increasing overall system power consumption.

When the TEST pin is again taken low after a test or programming session, the fuse check mode and sense currents are terminated.

Activation of the fuse check mode occurs with the first negative edge on the TMS pin after power up or if TMS is being held low during power up. The second positive edge on the TMS pin deactivates the fuse check mode. After deactivation, the fuse check mode remains inactive until another POR occurs. After each POR the fuse check mode has the potential to be activated.

The fuse check current flows only when the fuse check mode is active and the TMS pin is in a low state (see Figure 22). Therefore, the additional current flow can be prevented by holding the TMS pin high (default condition).

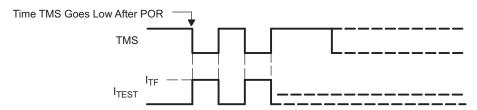


Figure 22. Fuse Check Mode Current

#### NOTE

The CODE and RAM data protection is ensured if the JTAG fuse is blown and the 256-bit bootloader access key is used. Also, see the Bootstrap Loader section for more information.



### **REVISION HISTORY**

Literature Number	Summary
SLAS439	PRODUCT PREVIEW release
SLAS439A	PRODUCTION DATA release
SLAS439B	Corrected instruction cycle time to 62.5ns, pg 1. Updated Figure 1, pg 12. Updated Figures 2 and 3, pg 13. $R_{Pull}$ unit corrected from $\Omega$ to $k\Omega$ , pg 15. MAX load current specification and Note 3 removed from "outputs" table, pg 16. MIN and MAX percentages for "calibrated DCO frequencies - tolerance over supply voltage VCC" corrected from 2.5% to 3% to match the specified frequency ranges., pg 22.
SLAS439C	MSP430x21x1T production data sheet release.  105°C characterization results added.
SLAS439D	Corrected Timer_A2 to Timer_A3 and added TACCR2 to Interrupt Flag column in "interrupt vector addresses", pg 6
SLAS439E	Changed T <sub>stg</sub> , Programmed device, to -40°C to 150°C in Absolute Maximum Ratings.  Corrected Test Conditions for OA <sub>HF</sub> row and and Duty Cycle row in Crystal Oscillator LFXT1, High-Frequency Mode.
SLAS439F	Changed T <sub>stg</sub> , Programmed device, to -55°C to 150°C in Absolute Maximum Ratings.





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### **PACKAGING INFORMATION**

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
MSP430F2101IDGV	ACTIVE	TVSOP	DGV	20	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	4F2101	Samples
MSP430F2101IDGVR	ACTIVE	TVSOP	DGV	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	4F2101	Samples
MSP430F2101IDW	ACTIVE	SOIC	DW	20	25	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	M430F2101	Samples
MSP430F2101IDWR	ACTIVE	SOIC	DW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	M430F2101	Samples
MSP430F2101IPW	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430F2101	Samples
MSP430F2101IPWR	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430F2101	Samples
MSP430F2101IRGER	ACTIVE	VQFN	RGE	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	M430F 2101	Samples
MSP430F2101IRGET	ACTIVE	VQFN	RGE	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	M430F 2101	Samples
MSP430F2101TDGV	ACTIVE	TVSOP	DGV	20	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	4F2101T	Samples
MSP430F2101TDGVR	ACTIVE	TVSOP	DGV	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	4F2101T	Samples
MSP430F2101TDW	ACTIVE	SOIC	DW	20	25	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 105	M430F2101T	Samples
MSP430F2101TDWR	ACTIVE	SOIC	DW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 105	M430F2101T	Samples
MSP430F2101TPW	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 105	4F2101T	Samples
MSP430F2101TPWR	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 105	4F2101T	Samples
MSP430F2101TRGER	ACTIVE	VQFN	RGE	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	M430F 2101T	Samples
MSP430F2101TRGET	ACTIVE	VQFN	RGE	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	M430F 2101T	Samples
MSP430F2111IDGV	ACTIVE	TVSOP	DGV	20	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	4F2111	Samples





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Orderable Device	Status	Package Type	-	Pins	Package	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	<b>Top-Side Markings</b>	Samples
	(1)		Drawing		Qty	(2)		(3)		(4)	
MSP430F2111IDGVR	ACTIVE	TVSOP	DGV	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	4F2111	Samples
MSP430F2111IDW	ACTIVE	SOIC	DW	20	25	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	M430F2111	Samples
MSP430F2111IDWR	ACTIVE	SOIC	DW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	M430F2111	Samples
MSP430F2111IPW	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430F2111	Samples
MSP430F2111IPWR	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430F2111	Samples
MSP430F2111IRGER	ACTIVE	VQFN	RGE	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	M430F 2111	Samples
MSP430F2111IRGET	ACTIVE	VQFN	RGE	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	M430F 2111	Samples
MSP430F2111TDGV	ACTIVE	TVSOP	DGV	20	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	4F2111T	Samples
MSP430F2111TDGVR	ACTIVE	TVSOP	DGV	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	4F2111T	Samples
MSP430F2111TDW	ACTIVE	SOIC	DW	20	25	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 105	M430F2111T	Samples
MSP430F2111TDWR	ACTIVE	SOIC	DW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 105	M430F2111T	Samples
MSP430F2111TPW	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 105	4F2111T	Samples
MSP430F2111TPWR	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 105	4F2111T	Samples
MSP430F2111TRGER	ACTIVE	VQFN	RGE	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	M430F 2111T	Samples
MSP430F2111TRGET	ACTIVE	VQFN	RGE	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	M430F 2111T	Samples
MSP430F2121IDGV	ACTIVE	TVSOP	DGV	20	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	4F2121	Samples
MSP430F2121IDGVR	ACTIVE	TVSOP	DGV	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	4F2121	Samples
MSP430F2121IDW	ACTIVE	SOIC	DW	20	25	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	M430F2121	Samples





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Orderable Device	Status	Package Type	-	Pins	_	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
	(1)		Drawing		Qty	(2)		(3)		(4)	
MSP430F2121IDWR	ACTIVE	SOIC	DW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	M430F2121	Samples
MSP430F2121IPW	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430F2121	Samples
MSP430F2121IPWR	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430F2121	Samples
MSP430F2121IRGER	ACTIVE	VQFN	RGE	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	M430F 2121	Samples
MSP430F2121IRGET	ACTIVE	VQFN	RGE	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	M430F 2121	Samples
MSP430F2121TDGV	ACTIVE	TVSOP	DGV	20	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	4F2121T	Samples
MSP430F2121TDGVR	ACTIVE	TVSOP	DGV	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	4F2121T	Samples
MSP430F2121TDW	ACTIVE	SOIC	DW	20	25	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 105	M430F2121T	Samples
MSP430F2121TDWR	ACTIVE	SOIC	DW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 105	M430F2121T	Samples
MSP430F2121TPW	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 105	4F2121T	Samples
MSP430F2121TPWR	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 105	4F2121T	Samples
MSP430F2121TRGER	ACTIVE	VQFN	RGE	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	M430F 2121T	Samples
MSP430F2121TRGET	ACTIVE	VQFN	RGE	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	M430F 2121T	Samples
MSP430F2131IDGV	ACTIVE	TVSOP	DGV	20	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	4F2131	Samples
MSP430F2131IDGVR	ACTIVE	TVSOP	DGV	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	4F2131	Samples
MSP430F2131IDW	ACTIVE	SOIC	DW	20	25	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	M430F2131	Samples
MSP430F2131IDWR	ACTIVE	SOIC	DW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	M430F2131	Samples
MSP430F2131IPW	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430F2131	Samples





11-Apr-2013

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
MSP430F2131IPWR	ACTIVE	TSSOP	PW	20	2000	(2) Green (RoHS & no Sb/Br)	CU NIPDAU	(3) Level-1-260C-UNLIM	-40 to 85	430F2131	Samples
MSP430F2131IRGER	ACTIVE	VQFN	RGE	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	M430F 2131	Samples
MSP430F2131IRGET	ACTIVE	VQFN	RGE	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	M430F 2131	Samples
MSP430F2131TDGV	ACTIVE	TVSOP	DGV	20	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	4F2131T	Samples
MSP430F2131TDGVR	ACTIVE	TVSOP	DGV	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	4F2131T	Samples
MSP430F2131TDW	ACTIVE	SOIC	DW	20	25	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 105	M430F2131T	Samples
MSP430F2131TDWR	ACTIVE	SOIC	DW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 105	M430F2131T	Samples
MSP430F2131TPW	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 105	4F2131T	Samples
MSP430F2131TPWR	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 105	4F2131T	Samples
MSP430F2131TRGER	ACTIVE	VQFN	RGE	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	M430F 2131T	Samples
MSP430F2131TRGET	ACTIVE	VQFN	RGE	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	M430F 2131T	Samples

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

TBD: The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.



### PACKAGE OPTION ADDENDUM

11-Apr-2013

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.

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In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

### **PACKAGE MATERIALS INFORMATION**

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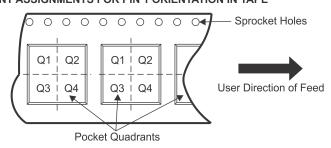
### TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



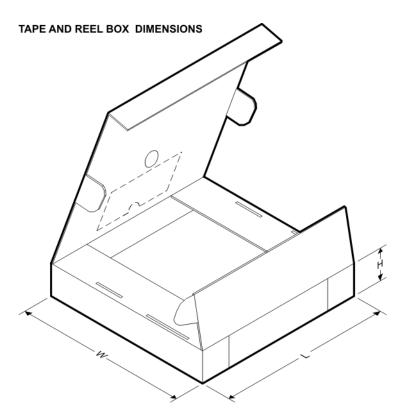
\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
MSP430F2101IDGVR	TVSOP	DGV	20	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
MSP430F2101IRGER	VQFN	RGE	24	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
MSP430F2101IRGET	VQFN	RGE	24	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
MSP430F2101TDGVR	TVSOP	DGV	20	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
MSP430F2101TPWR	TSSOP	PW	20	2000	330.0	16.4	6.95	7.1	1.6	8.0	16.0	Q1
MSP430F2111IDGVR	TVSOP	DGV	20	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
MSP430F2111IRGER	VQFN	RGE	24	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
MSP430F2111IRGET	VQFN	RGE	24	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
MSP430F2111TDGVR	TVSOP	DGV	20	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
MSP430F2111TRGER	VQFN	RGE	24	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
MSP430F2111TRGET	VQFN	RGE	24	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
MSP430F2121IDGVR	TVSOP	DGV	20	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
MSP430F2121IRGER	VQFN	RGE	24	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
MSP430F2121IRGET	VQFN	RGE	24	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
MSP430F2121TDGVR	TVSOP	DGV	20	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
MSP430F2121TPWR	TSSOP	PW	20	2000	330.0	16.4	6.95	7.1	1.6	8.0	16.0	Q1
MSP430F2121TRGER	VQFN	RGE	24	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
MSP430F2121TRGET	VQFN	RGE	24	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2

# **PACKAGE MATERIALS INFORMATION**

www.ti.com 11-Feb-2014

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
MSP430F2131IDGVR	TVSOP	DGV	20	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
MSP430F2131IRGER	VQFN	RGE	24	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
MSP430F2131IRGET	VQFN	RGE	24	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
MSP430F2131TDGVR	TVSOP	DGV	20	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
MSP430F2131TPWR	TSSOP	PW	20	2000	330.0	16.4	6.95	7.1	1.6	8.0	16.0	Q1
MSP430F2131TRGER	VQFN	RGE	24	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
MSP430F2131TRGET	VQFN	RGE	24	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2



### \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
MSP430F2101IDGVR	TVSOP	DGV	20	2000	367.0	367.0	35.0
MSP430F2101IRGER	VQFN	RGE	24	3000	367.0	367.0	35.0
MSP430F2101IRGET	VQFN	RGE	24	250	210.0	185.0	35.0
MSP430F2101TDGVR	TVSOP	DGV	20	2000	367.0	367.0	35.0
MSP430F2101TPWR	TSSOP	PW	20	2000	367.0	367.0	38.0
MSP430F2111IDGVR	TVSOP	DGV	20	2000	367.0	367.0	35.0
MSP430F2111IRGER	VQFN	RGE	24	3000	367.0	367.0	35.0
MSP430F2111IRGET	VQFN	RGE	24	250	210.0	185.0	35.0
MSP430F2111TDGVR	TVSOP	DGV	20	2000	367.0	367.0	35.0
MSP430F2111TRGER	VQFN	RGE	24	3000	367.0	367.0	35.0



# **PACKAGE MATERIALS INFORMATION**

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Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
MSP430F2111TRGET	VQFN	RGE	24	250	210.0	185.0	35.0
MSP430F2121IDGVR	TVSOP	DGV	20	2000	367.0	367.0	35.0
MSP430F2121IRGER	VQFN	RGE	24	3000	367.0	367.0	35.0
MSP430F2121IRGET	VQFN	RGE	24	250	210.0	185.0	35.0
MSP430F2121TDGVR	TVSOP	DGV	20	2000	367.0	367.0	35.0
MSP430F2121TPWR	TSSOP	PW	20	2000	367.0	367.0	38.0
MSP430F2121TRGER	VQFN	RGE	24	3000	367.0	367.0	35.0
MSP430F2121TRGET	VQFN	RGE	24	250	210.0	185.0	35.0
MSP430F2131IDGVR	TVSOP	DGV	20	2000	367.0	367.0	35.0
MSP430F2131IRGER	VQFN	RGE	24	3000	367.0	367.0	35.0
MSP430F2131IRGET	VQFN	RGE	24	250	210.0	185.0	35.0
MSP430F2131TDGVR	TVSOP	DGV	20	2000	367.0	367.0	35.0
MSP430F2131TPWR	TSSOP	PW	20	2000	367.0	367.0	38.0
MSP430F2131TRGER	VQFN	RGE	24	3000	367.0	367.0	35.0
MSP430F2131TRGET	VQFN	RGE	24	250	210.0	185.0	35.0

DW (R-PDSO-G20)

### PLASTIC SMALL OUTLINE



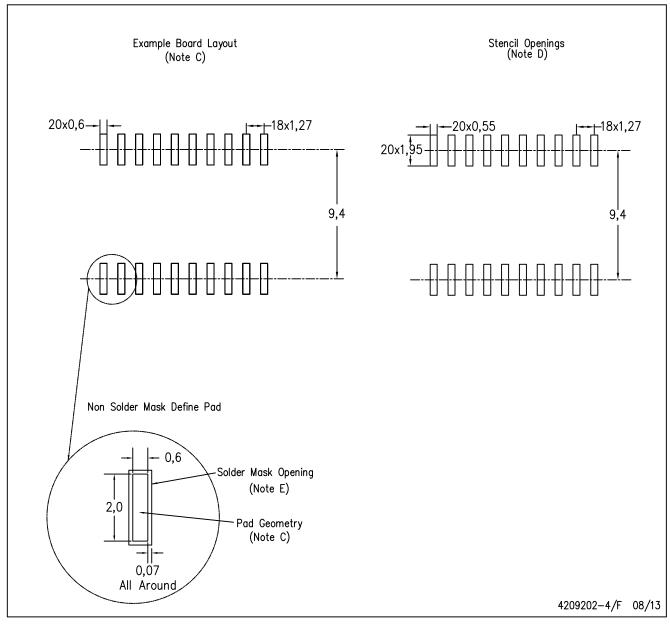
NOTES: A. All linear dimensions are in inches (millimeters). Dimensioning and tolerancing per ASME Y14.5M-1994.

- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).
- D. Falls within JEDEC MS-013 variation AC.



# DW (R-PDSO-G20)

## PLASTIC SMALL OUTLINE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Refer to IPC7351 for alternate board design.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



PW (R-PDSO-G20)

### PLASTIC SMALL OUTLINE



- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M—1994.
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.
- E. Falls within JEDEC MO-153



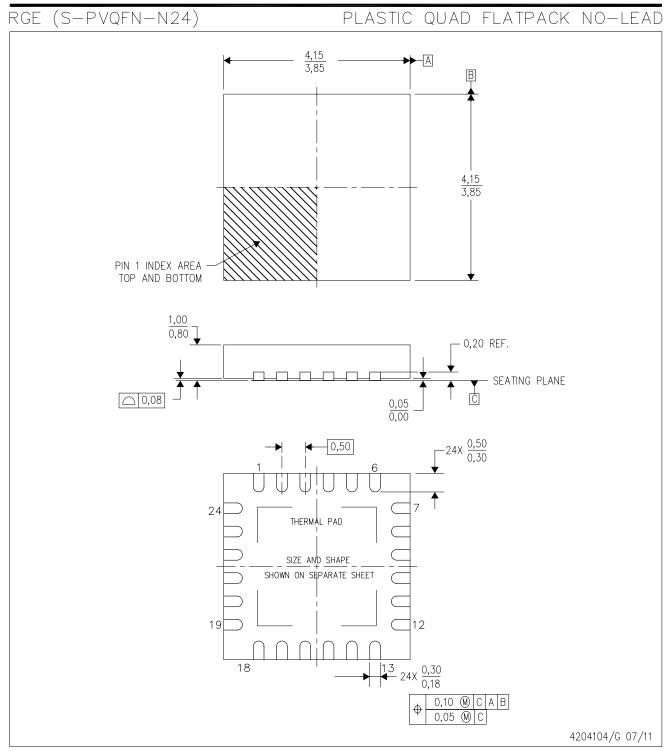
# PW (R-PDSO-G20)

## PLASTIC SMALL OUTLINE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate design.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.





- NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M—1994.
  - B. This drawing is subject to change without notice.
  - C. Quad Flatpack, No-Leads (QFN) package configuration.
  - D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
  - E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
  - F. Falls within JEDEC MO-220.



### RGE (S-PVQFN-N24)

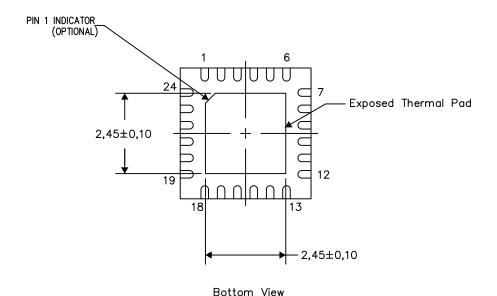
PLASTIC QUAD FLATPACK NO-LEAD

### THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No—Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Exposed Thermal Pad Dimensions

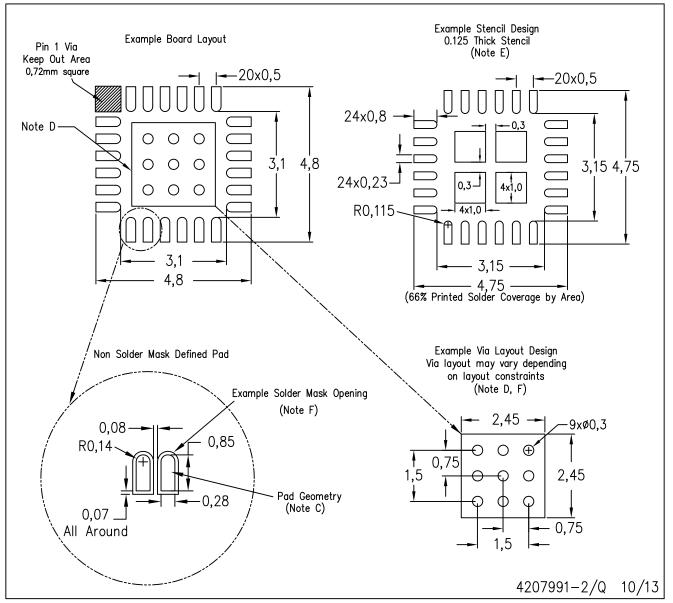
4206344-3/AE 10/13

NOTES: A. All linear dimensions are in millimeters



# RGE (S-PVQFN-N24)

### PLASTIC QUAD FLATPACK NO-LEAD



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat—Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <a href="http://www.ti.com">www.ti.com</a>.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- F. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in the thermal pad.



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